

Nonlinear Energy Harvesting

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N.i.P.S Laboratory
Noise in Physical Systems

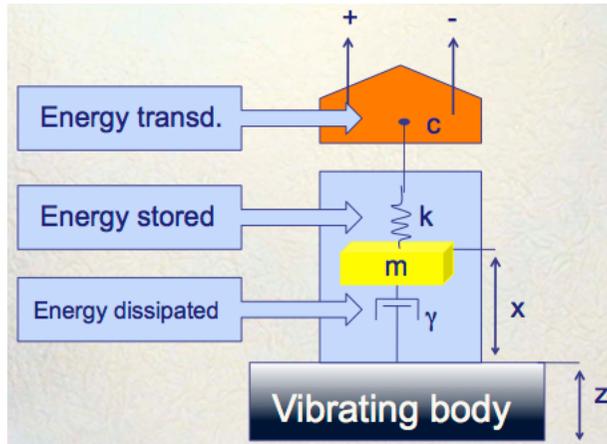
NANOPOWER

6 partners: Wuerzburg (Ger), ICN (Sp), VTT (Fi),
Univ Geneva (Ch), Unicam (It)
2.6 M€, 3 years, lead by NiPS
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ZEROPOWER

4 partners: UAB (Sp), Tyndall (Ir), Univ Glasgow (UK)
0.6 M€, 3 years, lead by NiPS

Vibrations energy harvesting



Dynamical model

$$m\ddot{x} = -\frac{dU(x)}{dx} - \gamma\dot{x} - c(x, V) + \xi_z$$

- Where:
- $U(x)$ Represents the Energy stored
 - $\gamma\dot{x}$ Represents the dissipative force
 - $c(x, V)$ Represents the reaction force due to the transduction mechanism
 - ξ_z Represents the input force

Vibrations energy harvesting

$$\left\{ \begin{array}{l} m\ddot{x} = -\frac{dU(x)}{dx} - \gamma\dot{x} - c(x,V) + \zeta_z \\ \dot{V} = F(\dot{x}, V) \end{array} \right.$$

Details depend on the physics...

Equations that link the vibration-induced displacement with the Voltage

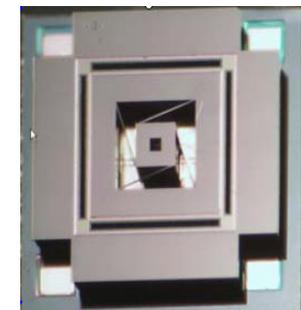
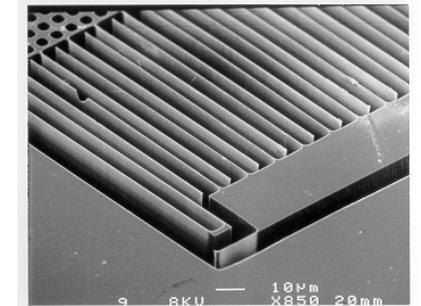
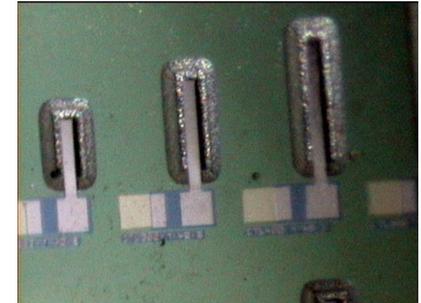
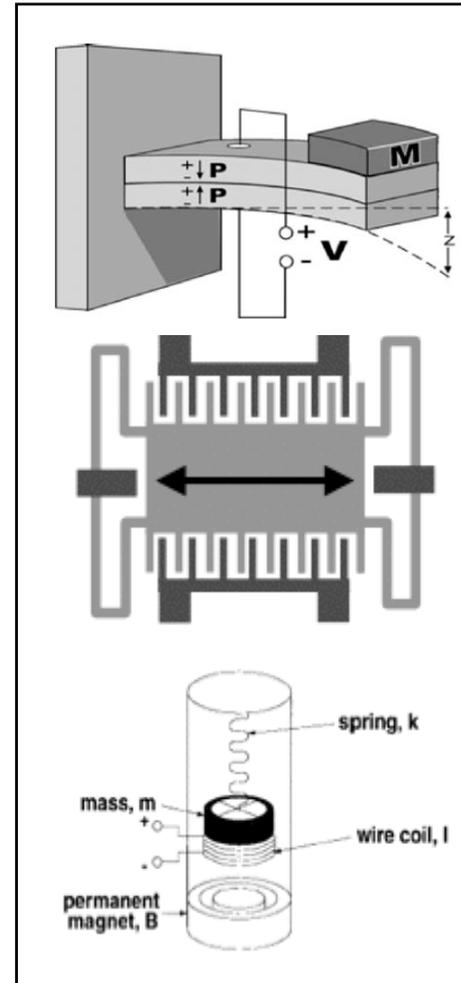
Vibrations energy harvesting

Transduction mechanisms

Piezoelectric: dynamical strain is converted into voltage difference.

Capacitive: geometrical variations induce voltage difference

Inductive: dynamical oscillations of magnets induce electric current in coils



Vibrations energy harvesting

Transduction mechanisms: focus on Piezo

Piezoelectric: dynamical strain is converted into voltage difference.

Type	Governing Equation	Practical Maximum	Theoretical Maximum
Piezoelectric	$u = \frac{\sigma_y^2 k^2}{2Y}$	17.7 mJ/cm ³	335 mJ/cm ³
Electrostatic	$u = \frac{1}{2} \epsilon E^2$	4 mJ/cm ³	44 mJ/cm ³
Electromagnetic	$u = \frac{B^2}{2\mu_0}$	4 mJ/cm ³	400 mJ/cm ³

Noise energy harvesting

Transduction mechanisms

Piezoelectric: dynamical strain is converted into voltage difference.

$$\left\{ \begin{array}{l} m\ddot{x} = -\frac{dU(x)}{dx} - \gamma\dot{x} - \underbrace{K_v V}_{\substack{\uparrow \\ \text{The Physics of piezo materials}}} + \zeta_z \\ \dot{V} = K_c \dot{x} - \underbrace{\frac{1}{\tau_p} V}_{\substack{\leftarrow \\ \text{The Physics of piezo materials}}} \end{array} \right.$$

That for a beam are:

$$K_v = \frac{K_{eff} d_{31} a}{2t_p k_1}$$

$$K_c = \frac{t_p d_{31} Y_p^E k_1}{a \epsilon_p}$$

The Physics of piezo materials

Focus on ζ_z

The random character of kinetic energy

ξ_z Represents the vibration (force)

Random vibrations / noise

Thermal noise (NOT POSSIBLE AT EQUILIBRIUM!!!)

Acoustic noise

Seismic noise

Ambient noise (wind, pressure fluctuations, ...)

Man made vibrations (human motion, machine vibrations,...)

All different for intensity, spectrum, statistics

How can we harvest them ?

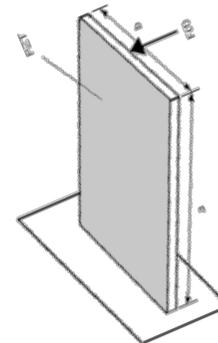
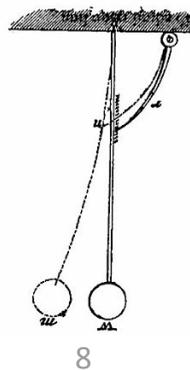
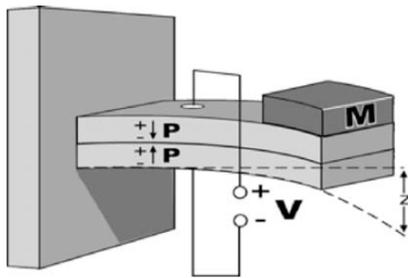
Noise energy harvesting

Linear systems

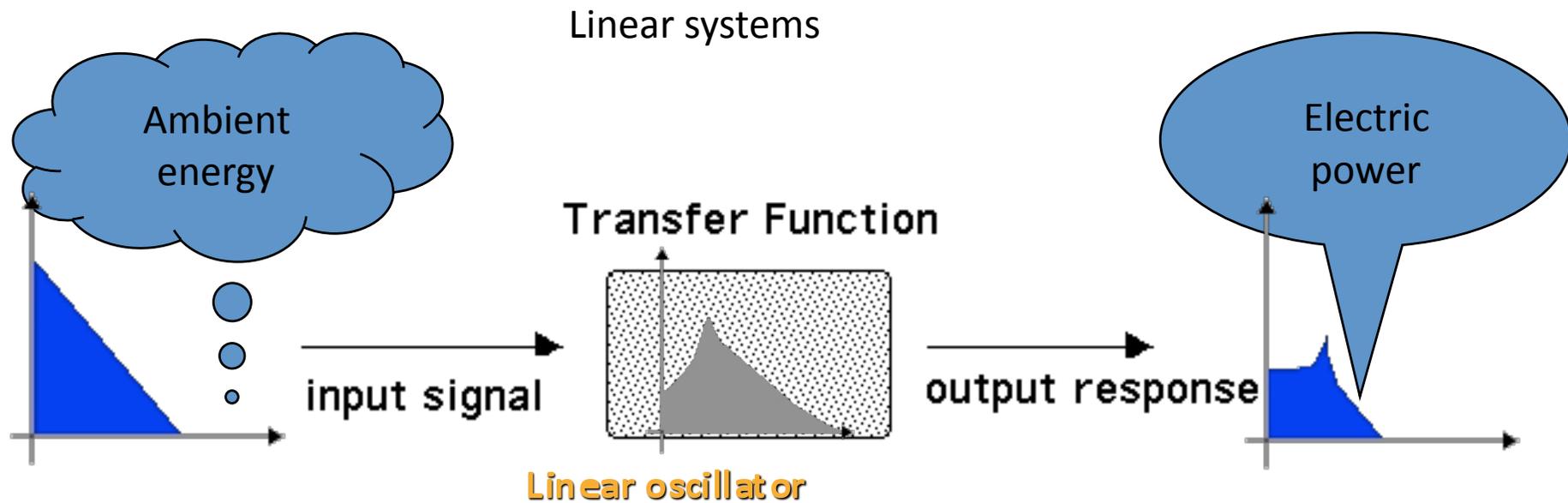
When $U(x) = \frac{1}{2} kx^2$ it is called a *linear system*

Linear systems have some interesting features... (and engineers like them most!!!)

- 1) There exist a simple math theory to solve the equations
- 2) They have a resonant behaviour (resonance frequency)
- 3) They can be “easily” realized with cantilevers and pendula



Vibrations energy harvesting



The transfer function is a math function of the frequency, in the complex domain, that can be used to represent the performance of a linear system

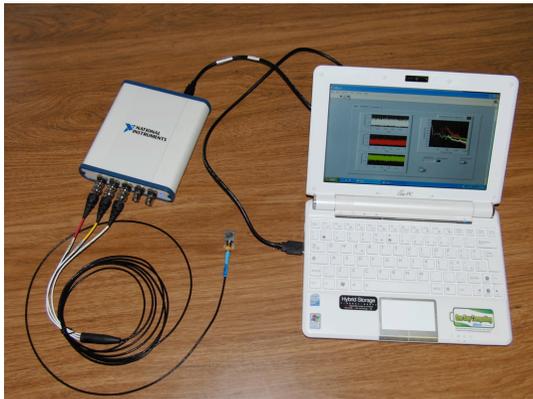
For a linear system the transfer function presents one or more peaks corresponding to the resonance frequencies and **thus it is efficient mainly when the incoming energy is abundant in that regions...**

This is a serious limitation when you want to build a small energy harvesting system...

Vibrations energy harvesting

For two main reasons...

- (1) the frequency spectrum of available vibrations instead of being sharply peaked at some frequency is usually very broad.
(See Igor Neri presentation at the workshop).



Accelerometer:

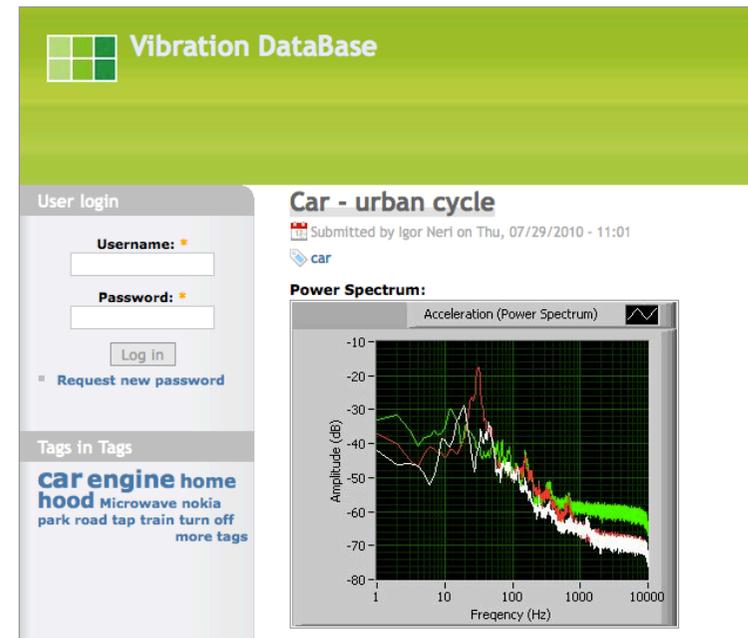
- Tri axial
- Bandwidth from 0.4Hz to 10kHz
- $\pm 50g$

DAQ:

- 102.4 kS/s five simultaneous channel
- 4 channels with software-selectable IEPE signal conditioning
- USB powered

Signal presentation:

- Description
- Power spectrum
- Statistical data
- Time series download (only for authorized users)

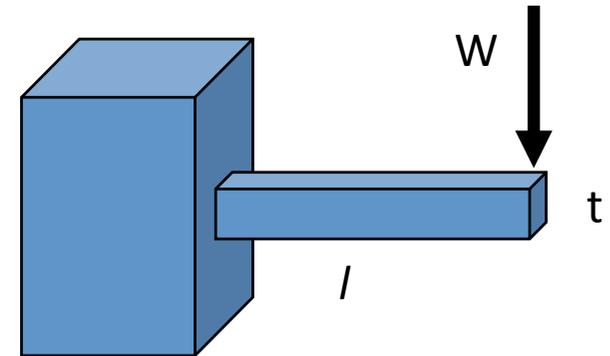


(2)

The frequency spectrum of available vibrations is particularly rich in energy in the low frequency part... and it is very difficult, if not impossible, to build small low-frequency resonant systems...

Resonant frequency $\sim [s^{-1}]$

- MEMS cantilever $100 \times 3 \times 0.1 \mu m^3$, $f_0=12$ kHz
- NEMS cantilever $0.1 \times 0.01 \times 0.01 \mu m^3$, $f_0=1.2$ GHz



$$f_0 = \frac{1}{2\pi} \sqrt{\frac{k}{m}} \quad \delta = \frac{Wl^3}{3EI} \quad k = \frac{W}{\delta} = \frac{3EI}{l^3}$$

$$f_0 = \frac{1}{2\pi} \sqrt{\frac{3EI}{Ml^3}} = \frac{1}{2\pi} \sqrt{\frac{Ewt^3}{4Ml^3}} = \frac{t}{4\pi l^2} \sqrt{\frac{E}{\rho}}$$

Let's look an example...

Description of the resonator design

The resonator design is a square shaped block of single crystal silicon with dimensions of $320 \times 320 \times 28 \text{ um}^3$ (design H1). Its main resonance mode is the so called square extensional (SE) resonance, which is characterized by its zoom-in/zoom-out oscillation. The resonance is excited by a piezoelectric AlN thin film on top of the resonator block. The electrically conductive (p-doped) silicon block acts as the bottom electrode, and a molybdenum thin film has been patterned to provide the top electrode. See reference [1] for a general description of the SE resonator. Reference [2] discusses piezoelectric excitation of the SE resonance mode.

Figure 3 shows how the resonator is recommended to be connected.

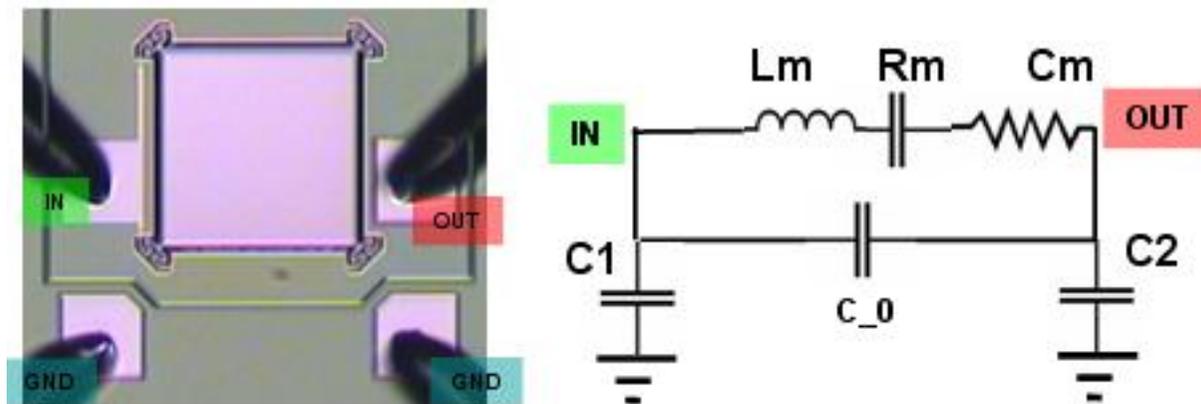
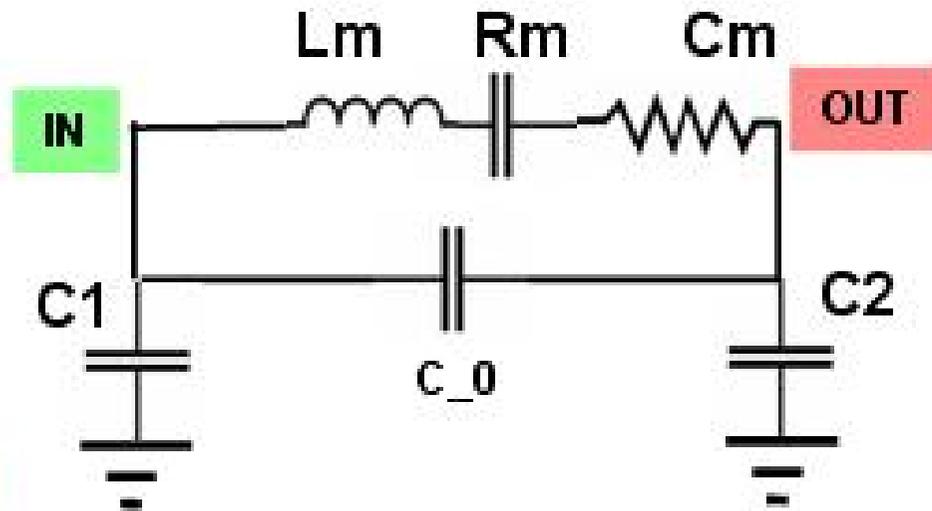


Figure 3: Electrical connection of the resonator.

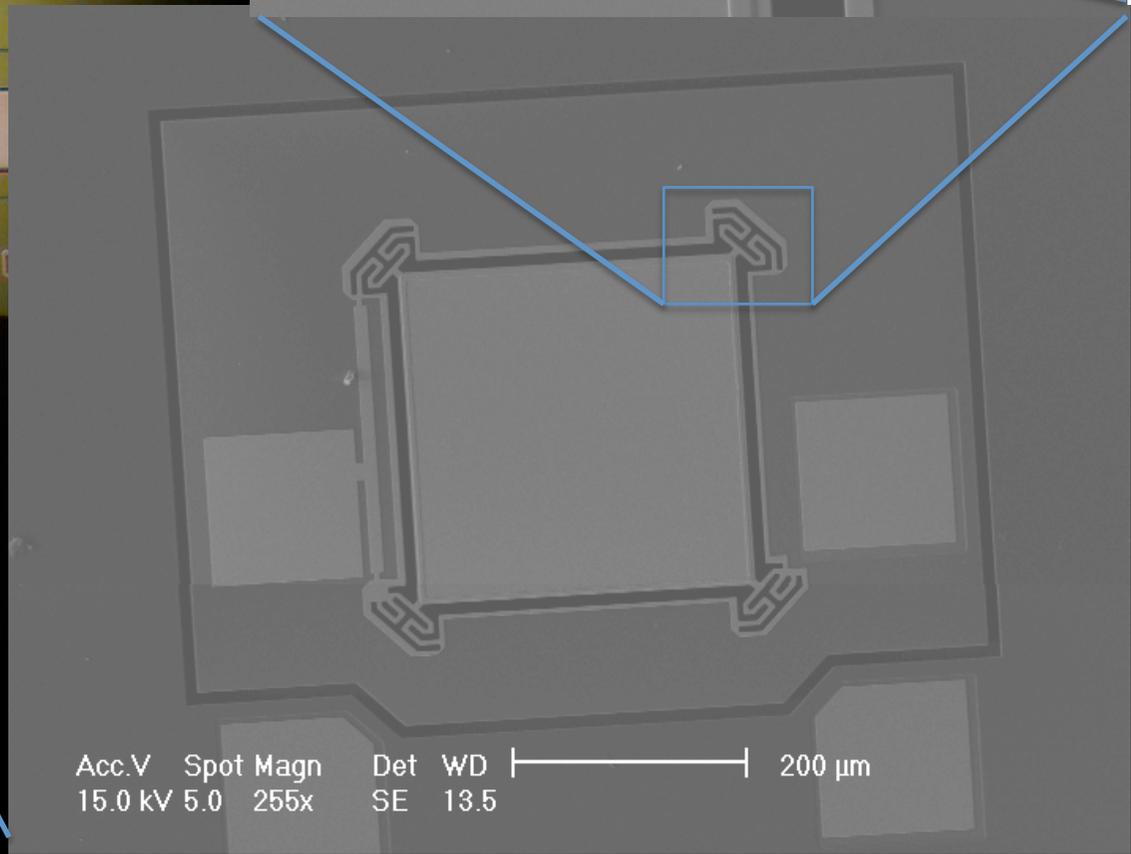
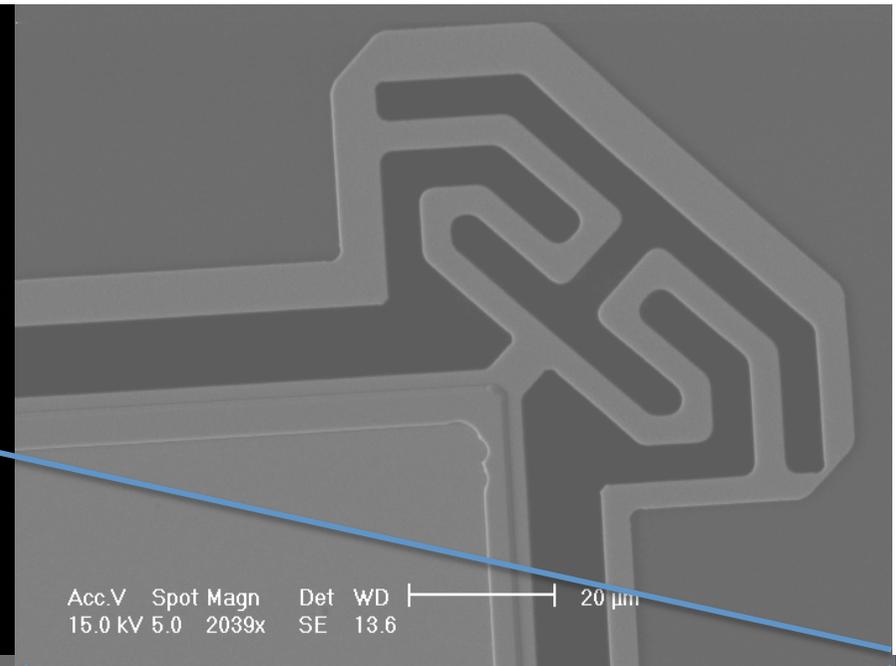
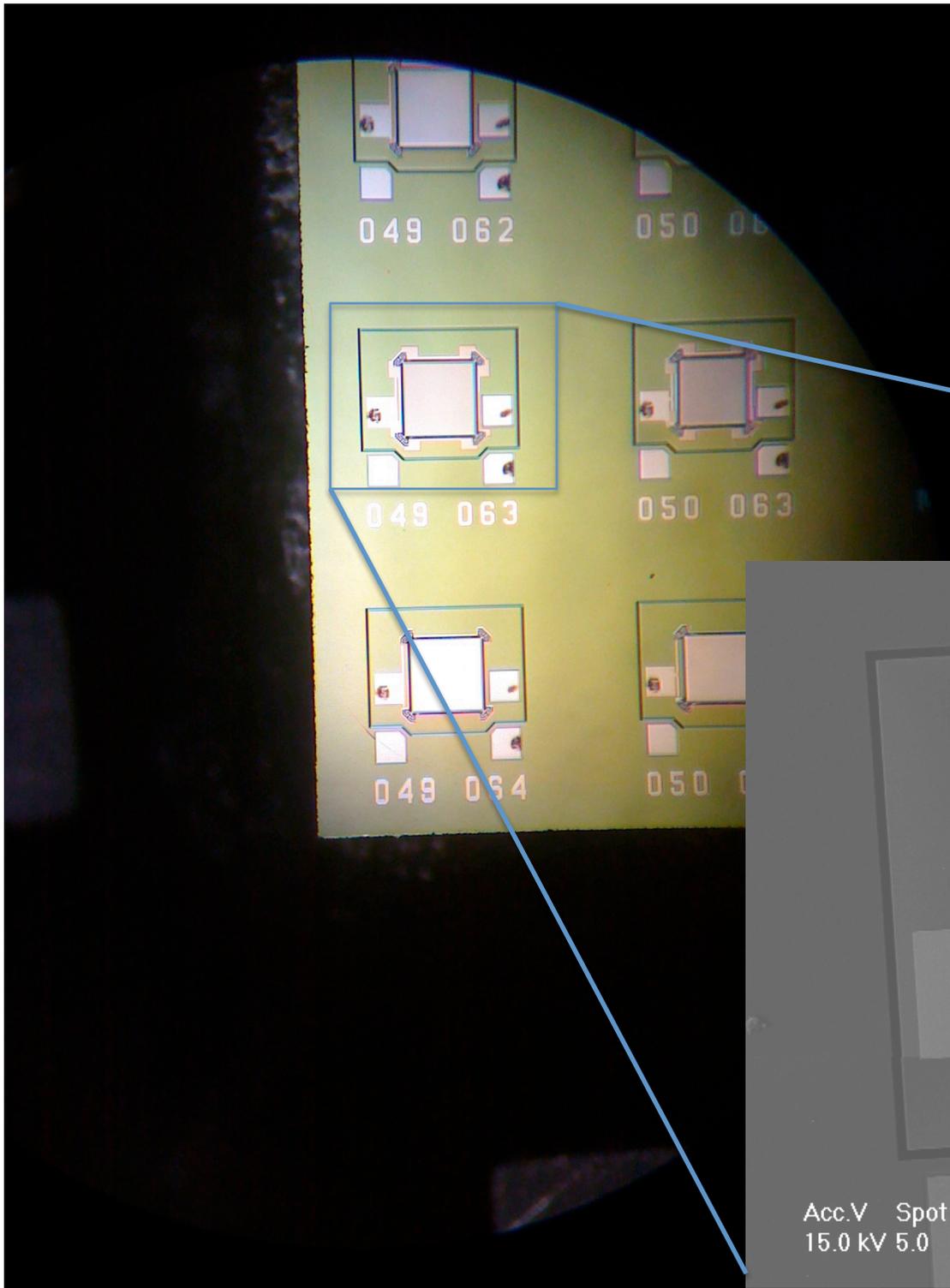
References

- [1] V. Kaajakari et al., "Square-Extensional Mode Single-Crystal Silicon Micromechanical Resonator for Low-Phase-Noise Oscillator Applications," *IEEE Electron Device Letters* 25, no. 4 (4, 2004): 173-175.
- [2] A. Jaakkola et al., "Piezoelectrically transduced Single-Crystal-Silicon Plate Resonators," in *IEEE Ultrasonics Symposium* (presented at the IEEE Ultrasonics Symposium, Beijing, China, 2008), 2181 – 2184

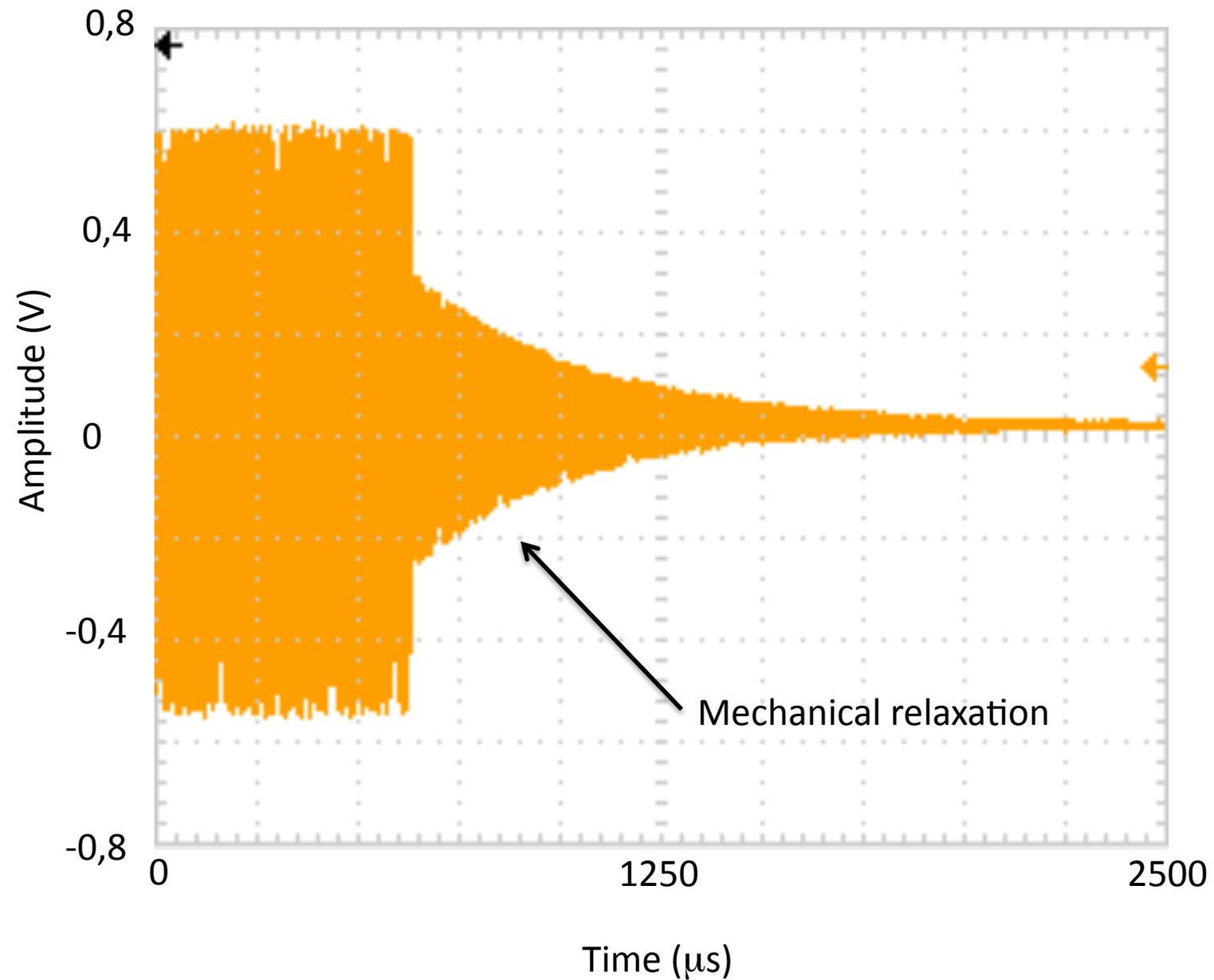


parameter	value	unit
R_m	128.9606	Ohm
C_m	1.01E-14	F
L_m	0.013691	H
C_0	3.00E-11	F
f_0	13.5	MHz
Q	9000	1
k_{2eff}	0.04	%
C_1	<1e-12	F
C_2	<1e-12	F

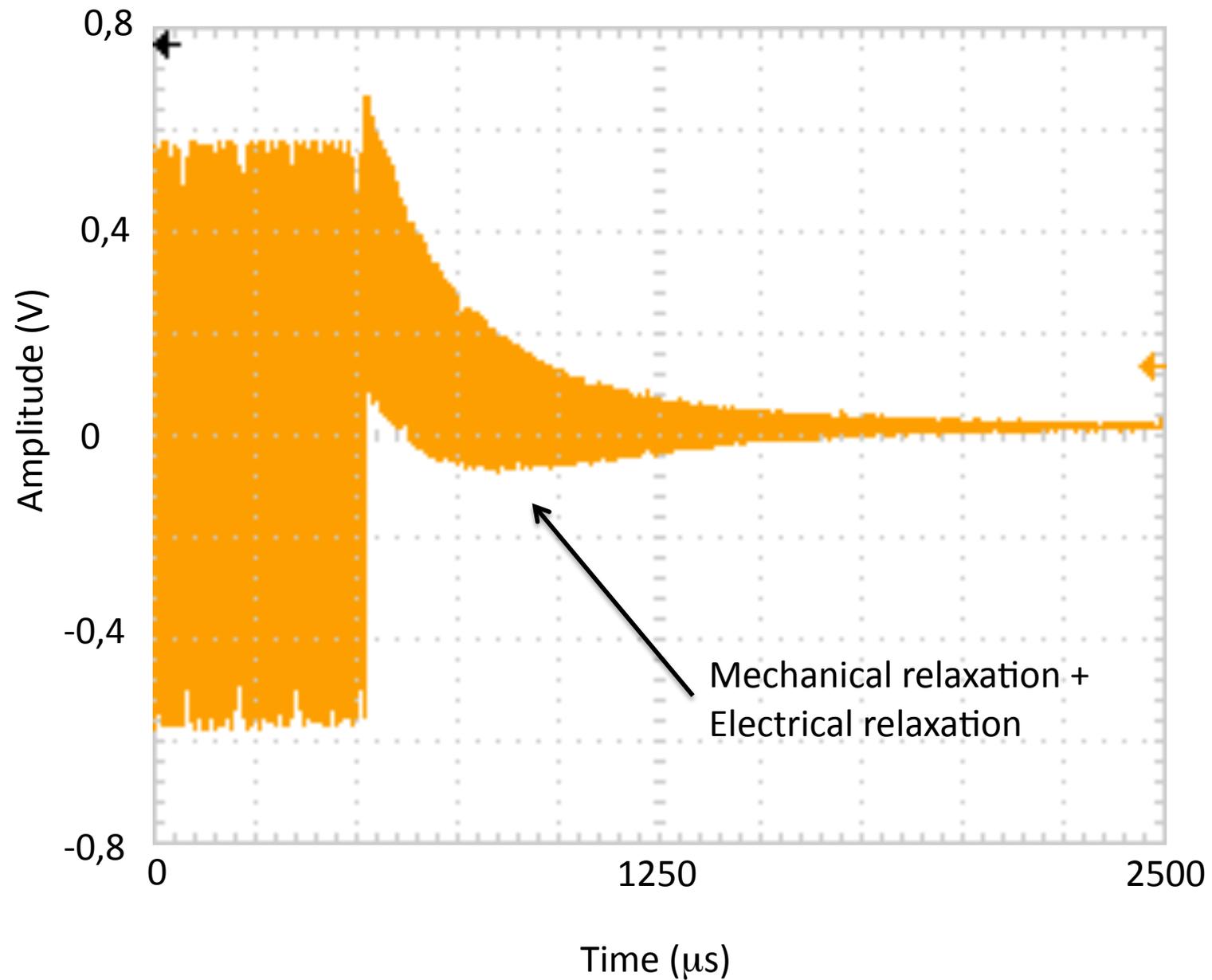
Table 1: equivalent circuit parameters (and their derivatives).



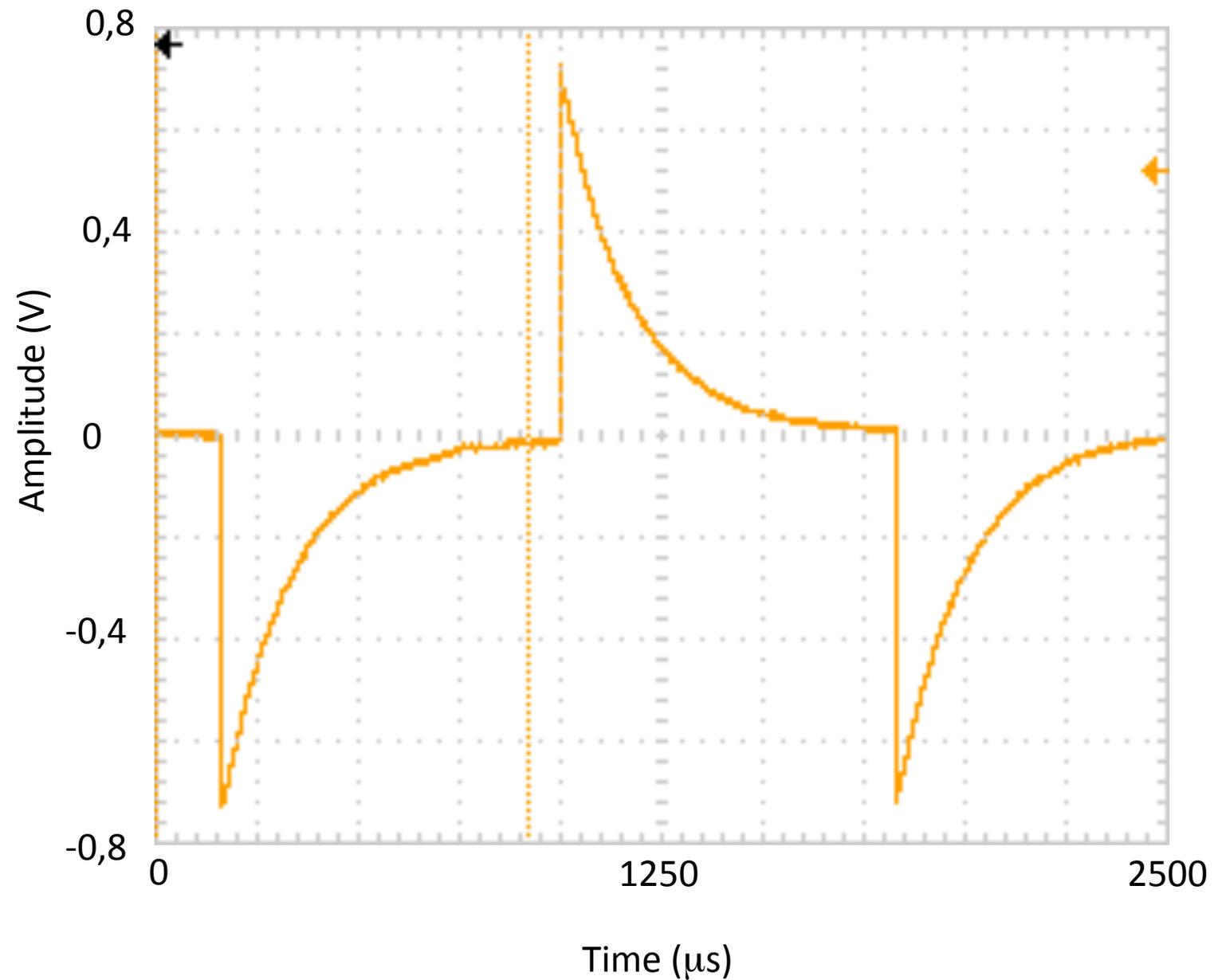
Voltage sinusoidal excitation switched off at zero



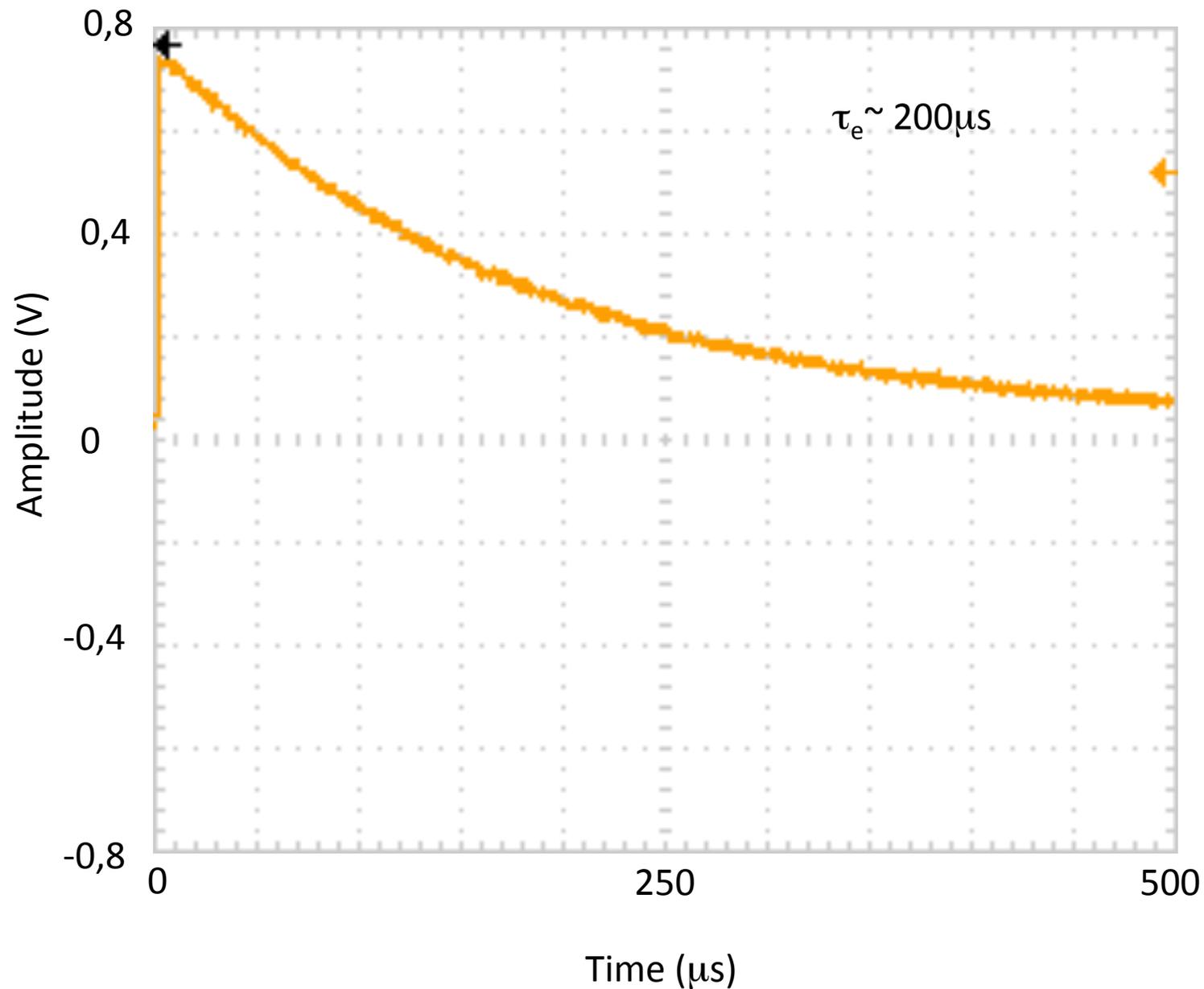
Voltage sinusoidal excitation switched off at max



Square wave excitation



Electrical relaxation time (τ_e)



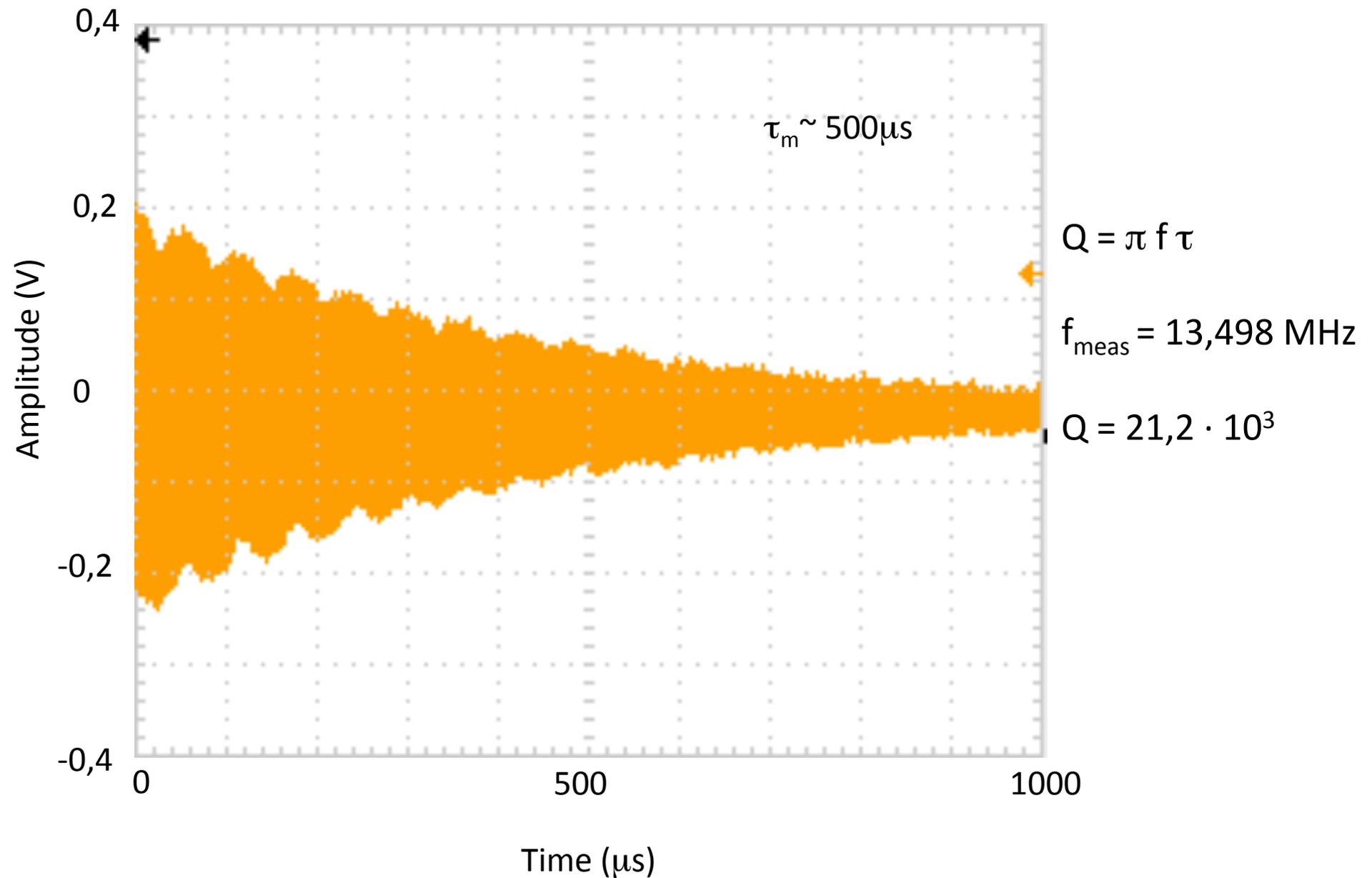
$$\tau_e = RC$$

$$C_{\text{meas.}} = 24 \text{ pF}$$

$$R_p = 8 \text{ M}\Omega$$

Mechanical relaxation time (τ_m) in vacuum

(open circuit)



From the model for a linear oscillator:

The voltage transfer function is:

$$|H(\omega)| = \frac{1}{m} \frac{\omega k_c}{\sqrt{\omega^2 \left(\frac{k + \frac{\gamma}{\tau} + k_c k_v}{m} - \omega^2 \right)^2 + \left(\left(\frac{\gamma}{m} + \frac{1}{\tau} \right) \omega^2 - \frac{k}{m\tau} \right)^2}} \quad \text{or considering: } \omega_0 = \sqrt{\frac{k + \frac{\gamma}{\tau} + k_c k_v}{m}}$$

and:

$$\omega_1 = \sqrt{\frac{k}{\gamma\tau + m}}$$



$$|H(\omega)| = \frac{1}{m} \frac{\omega k_c}{\sqrt{\omega^2 (\omega^2 - \omega_0^2)^2 + \left(\frac{\gamma\tau + m}{m\tau} \right)^2 (\omega^2 - \omega_1^2)^2}}$$

if $\omega^2 (\omega^2 - \omega_0^2)^2 > \left(\frac{\gamma\tau + m}{m\tau} \right)^2 (\omega^2 - \omega_1^2)^2$

the resonance frequency is $\omega_0 = \sqrt{\frac{k + \frac{\gamma}{\tau} + k_c k_v}{m}}$

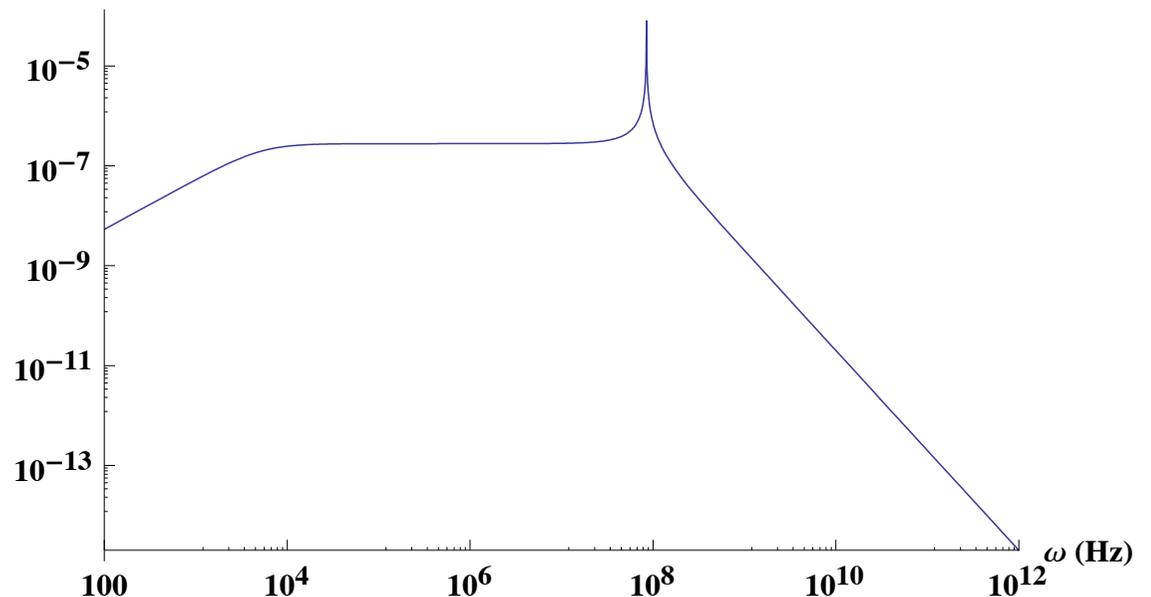
where $|H(\omega)|_{\max} = \frac{\omega_0 k_c \tau}{(\gamma\tau + m)(\omega_0^2 - \omega_1^2)^2}$

if $\omega^2 (\omega^2 - \omega_0^2)^2 < \left(\frac{\gamma\tau + m}{m\tau} \right)^2 (\omega^2 - \omega_1^2)^2$

the resonance frequency is $\omega_1 = \sqrt{\frac{k}{\gamma\tau + m}}$

where $|H(\omega)|_{\max} = \frac{k_c}{m|\omega_1^2 - \omega_0^2|}$

Units/ $\sqrt{\text{Hz}}$



The analytic result for the Q is:

$$Q = \frac{\omega_r}{\Delta\omega} \quad \omega_r \text{ is the resonance frequency and } \Delta\omega \text{ is the bandwidth (full width when the output voltage is } \frac{V_{\max}}{\sqrt{2}} \text{)}$$

$$\begin{aligned} \text{Quality Factor} = & \left\{ \left(3 k m \tau^2 \sqrt{\left(2 m^2 + 2 (k + kc kv) m \tau^2 - \gamma^2 \tau^2 + \right. \right. \right. \\ & \left. \left. \sqrt{m^4 + 2 (k - 2 kc kv) m^3 \tau^2 - 4 (k + kc kv) m \gamma^2 \tau^4 + \gamma^4 \tau^4 + m^2 \tau^2 (-\gamma^2 - 6 kc kv \gamma \tau + (k + kc kv)^2 \tau^2)} \right) \right) / \left(\sqrt{\left(2 m^6 + 6 (k \right. \right. \right. \\ & \left. \left. - 2 kc kv) m^5 \tau^2 + 2 \gamma^6 \tau^6 - 2 \gamma^4 \tau^4 \right. \right. \\ & \left. \left. \sqrt{m^4 + 2 (k - 2 kc kv) m^3 \tau^2 - 4 (k + kc kv) m \gamma^2 \tau^4 + \gamma^4 \tau^4 + m^2 \tau^2 (-\gamma^2 - 6 kc kv \gamma \tau + (k + kc kv)^2 \tau^2)} + 4 (k + kc kv) m \gamma^2 \right. \right. \\ & \left. \left. \tau^4 \left(-3 \gamma^2 \tau^2 + 2 \sqrt{m^4 + 2 (k - 2 kc kv) m^3 \tau^2 - 4 (k + kc kv) m \gamma^2 \tau^4 + \gamma^4 \tau^4 + m^2 \tau^2 (-\gamma^2 - 6 kc kv \gamma \tau + (k + kc kv)^2 \tau^2)} \right) \right) - \right. \\ & \left. m^4 \left(3 \gamma^2 \tau^2 + 18 kc kv \gamma \tau^3 - 6 k^2 \tau^4 + 6 k kc kv \tau^4 - 15 kc^2 kv^2 \tau^4 + 2 \right. \right. \\ & \left. \left. \sqrt{m^4 + 2 (k - 2 kc kv) m^3 \tau^2 - 4 (k + kc kv) m \gamma^2 \tau^4 + \gamma^4 \tau^4 + m^2 \tau^2 (-\gamma^2 - 6 kc kv \gamma \tau + (k + kc kv)^2 \tau^2)} \right) + m^2 \tau^2 \left(-3 \gamma^4 \right. \right. \\ & \left. \left. \tau^2 - 18 kc kv \gamma^3 \tau^3 + 12 kc kv \gamma \tau \right. \right. \\ & \left. \left. \sqrt{m^4 + 2 (k - 2 kc kv) m^3 \tau^2 - 4 (k + kc kv) m \gamma^2 \tau^4 + \gamma^4 \tau^4 + m^2 \tau^2 (-\gamma^2 - 6 kc kv \gamma \tau + (k + kc kv)^2 \tau^2)} - 2 (k + kc kv)^2 \right. \right. \\ & \left. \left. \tau^2 \sqrt{m^4 + 2 (k - 2 kc kv) m^3 \tau^2 - 4 (k + kc kv) m \gamma^2 \tau^4 + \gamma^4 \tau^4 + m^2 \tau^2 (-\gamma^2 - 6 kc kv \gamma \tau + (k + kc kv)^2 \tau^2)} + \gamma^2 \left(15 k^2 \tau^4 \right. \right. \\ & \left. \left. + 30 k kc kv \tau^4 + 15 kc^2 kv^2 \tau^4 + 2 \right. \right. \\ & \left. \left. \sqrt{m^4 + 2 (k - 2 kc kv) m^3 \tau^2 - 4 (k + kc kv) m \gamma^2 \tau^4 + \gamma^4 \tau^4 + m^2 \tau^2 (-\gamma^2 - 6 kc kv \gamma \tau + (k + kc kv)^2 \tau^2)} \right) \right) + 2 m^3 \tau^2 \left(\right. \\ & \left. k^3 \tau^4 + 3 k^2 kc kv \tau^4 + k \left(6 \gamma^2 \tau^2 + 18 kc kv \gamma \tau^3 + 3 kc^2 kv^2 \tau^4 - 2 \right. \right. \\ & \left. \left. \sqrt{m^4 + 2 (k - 2 kc kv) m^3 \tau^2 - 4 (k + kc kv) m \gamma^2 \tau^4 + \gamma^4 \tau^4 + m^2 \tau^2 (-\gamma^2 - 6 kc kv \gamma \tau + (k + kc kv)^2 \tau^2)} \right) + kc kv \left(-3 \right. \right. \\ & \left. \left. \gamma^2 \tau^2 + 18 kc kv \gamma \tau^3 + kc^2 kv^2 \tau^4 + 4 \right. \right. \\ & \left. \left. \sqrt{m^4 + 2 (k - 2 kc kv) m^3 \tau^2 - 4 (k + kc kv) m \gamma^2 \tau^4 + \gamma^4 \tau^4 + m^2 \tau^2 (-\gamma^2 - 6 kc kv \gamma \tau + (k + kc kv)^2 \tau^2)} \right) \right) \right) \right\} \end{aligned}$$

Considering:

$$m=10^{-6} \text{ g}$$

$$k=7.195 \cdot 10^9 \text{ N/m}$$

$$k_v=2 \text{ N/V}$$

$$k_c=2000 \text{ V/m}$$

$$R_0=8 \cdot 10^6 \text{ } \Omega$$

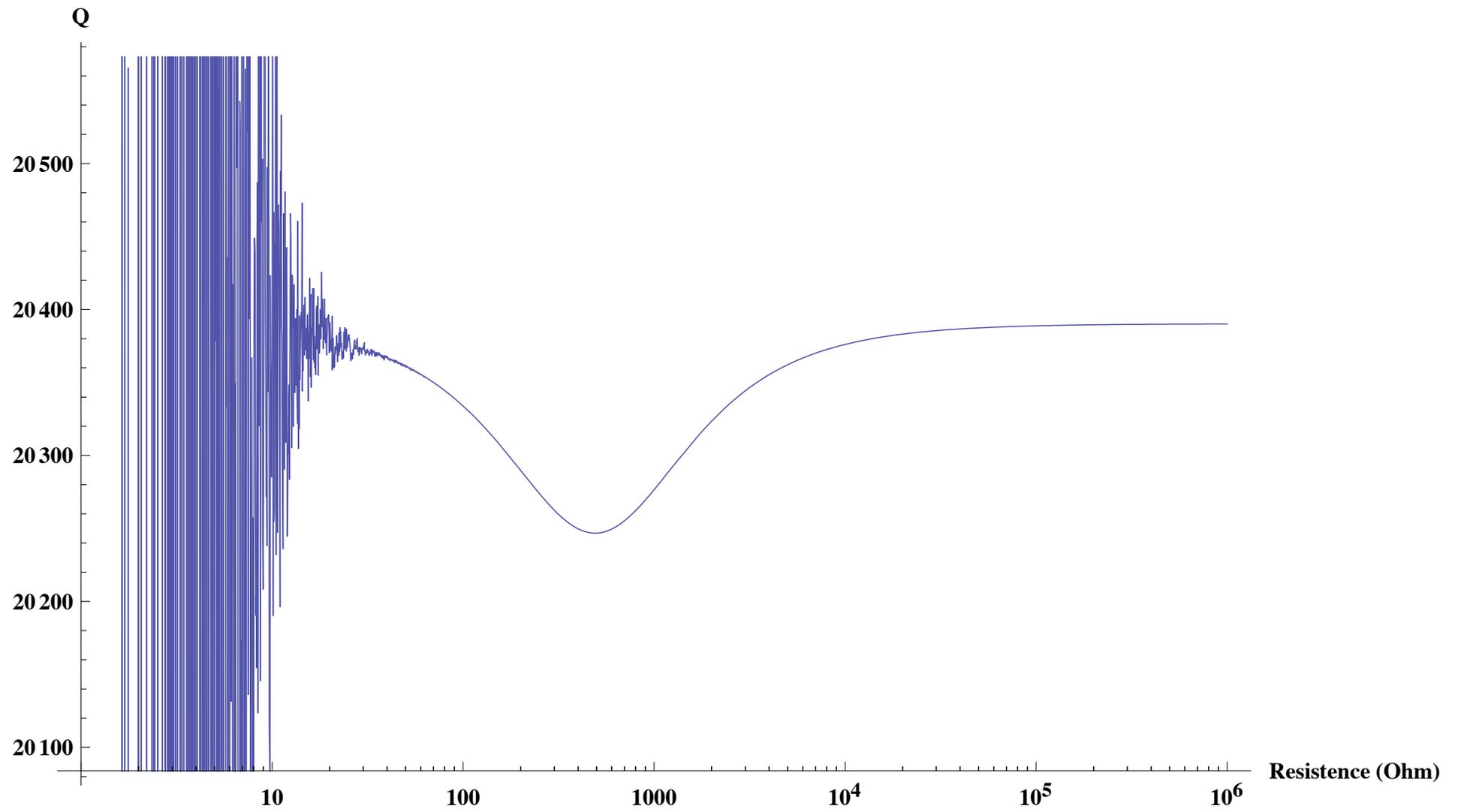
$$C_0=24 \cdot 10^{-12} \text{ F}$$

$$\gamma=m \cdot 4.16 \cdot 10^3 \text{ s}^{-1}$$

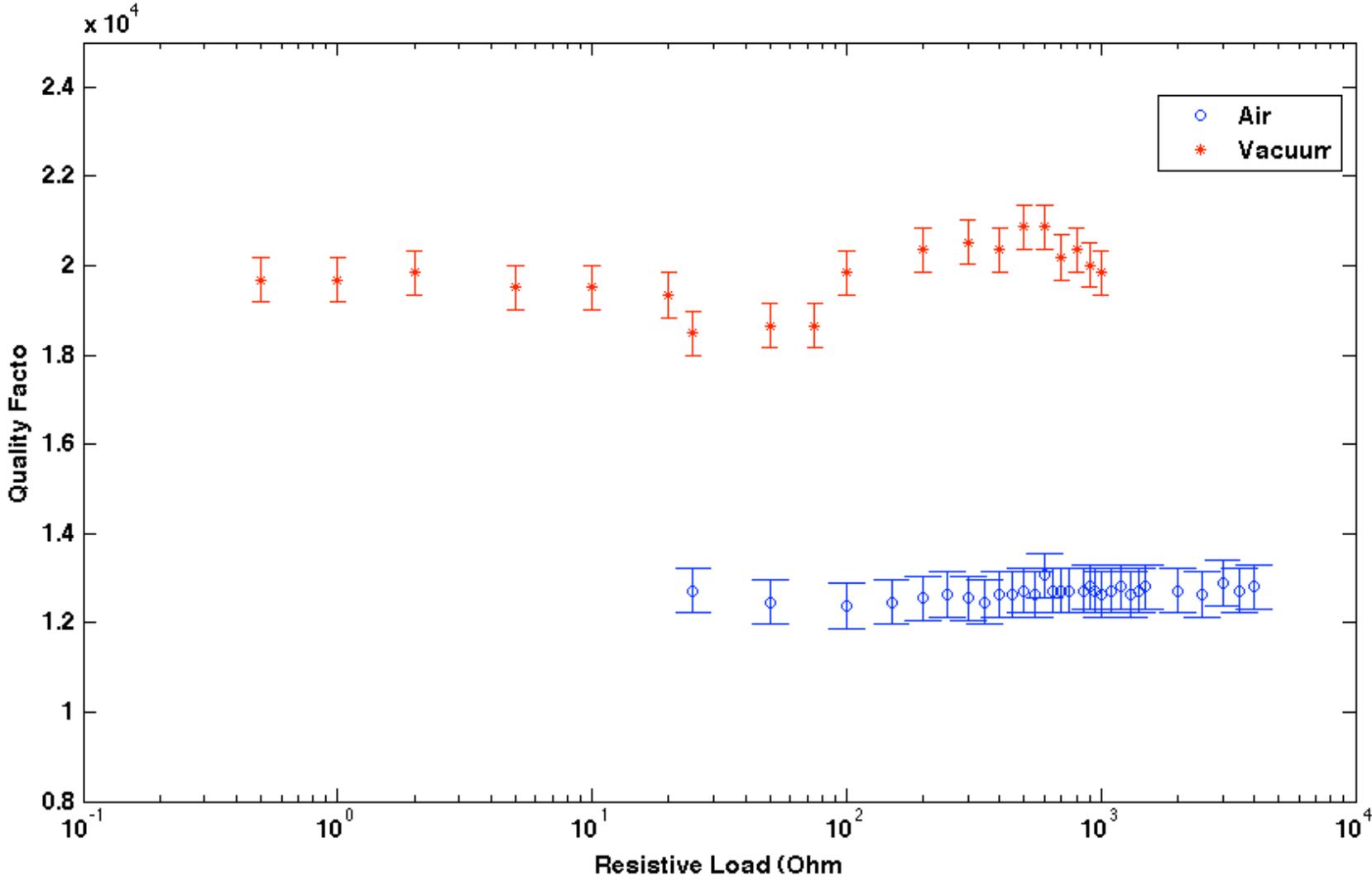
$$K_v = \frac{K_{eff} d_{31} a}{2t_p k_1}$$

$$K_c = \frac{t_p d_{31} Y_p^E k_1}{a \varepsilon_p}$$

Quality Factor as a function of the Load resistance



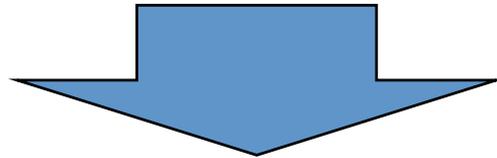
Quality Factor vs Resistive load



Vibrations energy harvesting

Which list for the perfect vibration harvester

- 1) Capable of harvesting energy on a broad-band
- 2) No need for frequency tuning
- 3) Capable of harvesting energy at low frequency



- 1) Non-resonant system
- 2) "Transfer function" with wide frequency resp.
- 3) Low frequency operated

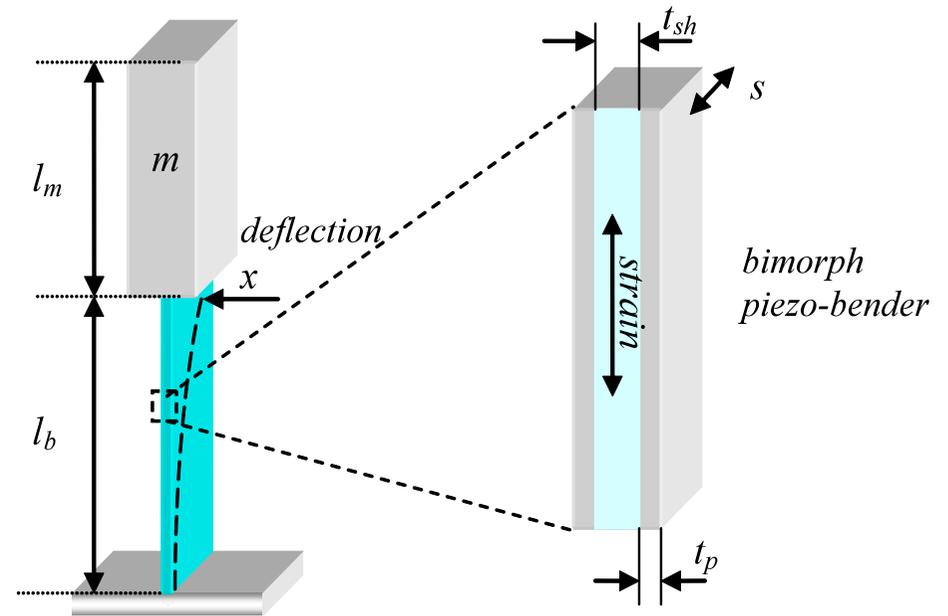
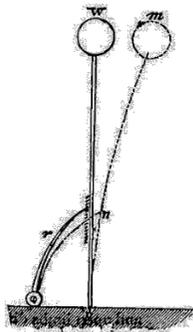
Noise energy harvesting

NON-Linear mechanical oscillators

- 1) Non-resonant system
- 2) Wide frequency resp.
- 3) Low frequency operated

Example...

Inverted pendulum



F. Cottone, PhD Thesis, Perugia 2007

Statistics

- “1D” Statistics: (2nd Order Cumulants, 1st Order Spectra)

– Correlation: $C_{xy}(t) = \int_{-\infty}^{\infty} x(\tau) y(t + \tau) d\tau \Leftrightarrow X(f) Y^*(f) = S_{xy}(f)$

– Power Spectral Density: $C_{2x}(t) \Leftrightarrow X(f) X^*(f) = S_{2x}(f)$

– Coherence: $C_{xy}(f) = \frac{S_{xy}(f)}{\sqrt{S_{2x}(f) S_{2y}(f)}}$

- Tells us power and phase coherence at a given frequency

Statistics (more complicated...)

- “2D” Statistics: (3rd Order Cumulants, 2nd Order Spectra)

– Bicumulant:

$$C_{xyz}(t, t') = \int_{-\infty}^{\infty} x(\tau) y(t + \tau) z(t' + \tau) d\tau \Leftrightarrow X(f_1) Y(f_2) Z^*(f_1 + f_2) = S_{xyz}(f_1, f_2)$$

– Bispectral Density: $C_{3x}(t) \Leftrightarrow X(f_1) X(f_2) X^*(f_1 + f_2) = S_{3x}(f_1, f_2)$

$$S_{3x}(f_1, f_2) = \int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} C_{3x}(m, n) e^{2\pi i(f_1 m + f_2 n)} dm dn$$

– Bicoherence: $\mathbf{c}_{xyz}(f) = \frac{S_{xyz}(f_1, f_2)}{\sqrt{S_{xx}(f_1)} \sqrt{S_{yy}(f_2)} \sqrt{S_{zz}(f_1 + f_2)}}$

- Tells us power and phase coherence at a coupled frequency

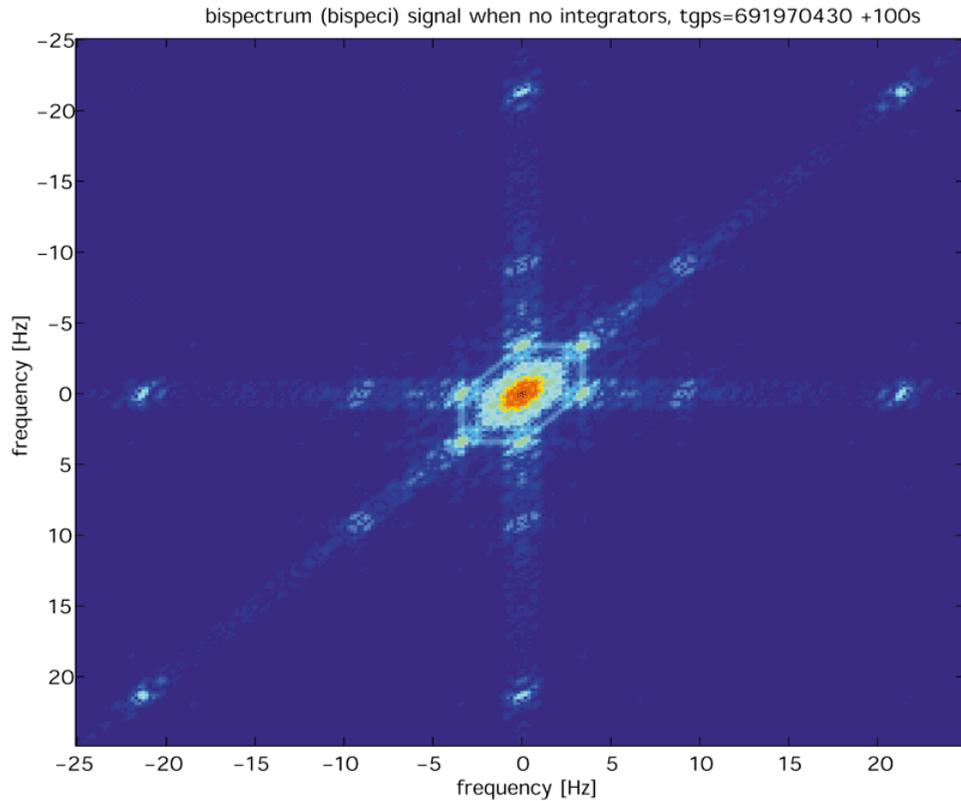
Statistics (more complicated...)

The Spectrogram (STFT square modulus):

$$S_x(t, \nu) = \left| \int_{-\infty}^{+\infty} x(\tau) h^*(\tau - t) e^{-i2\pi\nu\tau} d\tau \right|^2$$

Represents the signal energy in the time-frequency domain centred in (t, ν) .

- To analyze the system linearity bispectrum and bicoherence need to be taken into account:
- If $S_{3x} = 0$ the process is Gaussian and linear
- If $S_{3x} \neq 0$ the process is not Gaussian and
 - if c_{3x} is constant - the process is linear
 - if c_{3x} is not constant - the process is not linear

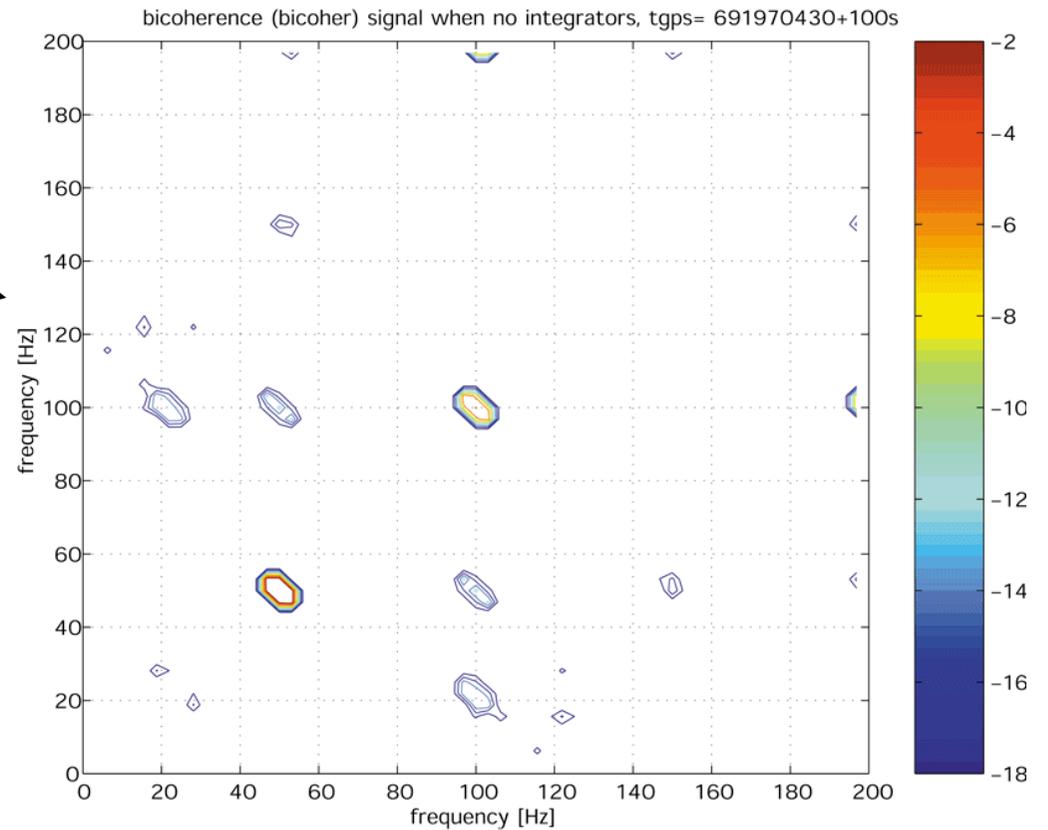


Bispectrum

Low frequency noise coupled at higher frequencies

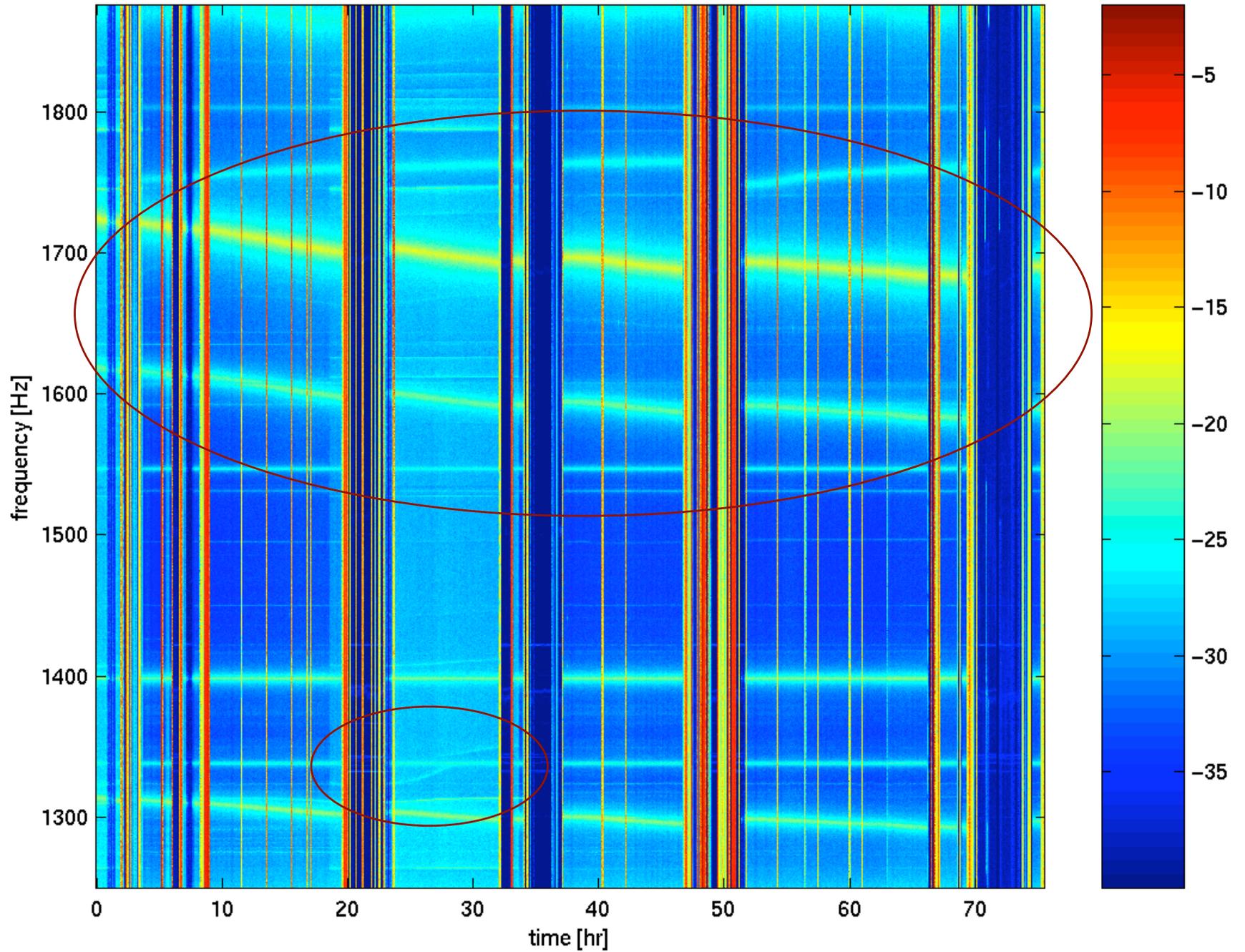
Bicoherence

A nonlinearity of the 50 Hz with its armonics is observed. There is present a big coupling between the 20 Hz and the 100 Hz and a smaller one between the 20 and 30 Hz.



Spectrogram:

start time: GPS=710517543, local=12 Jul 2002 15:58:54



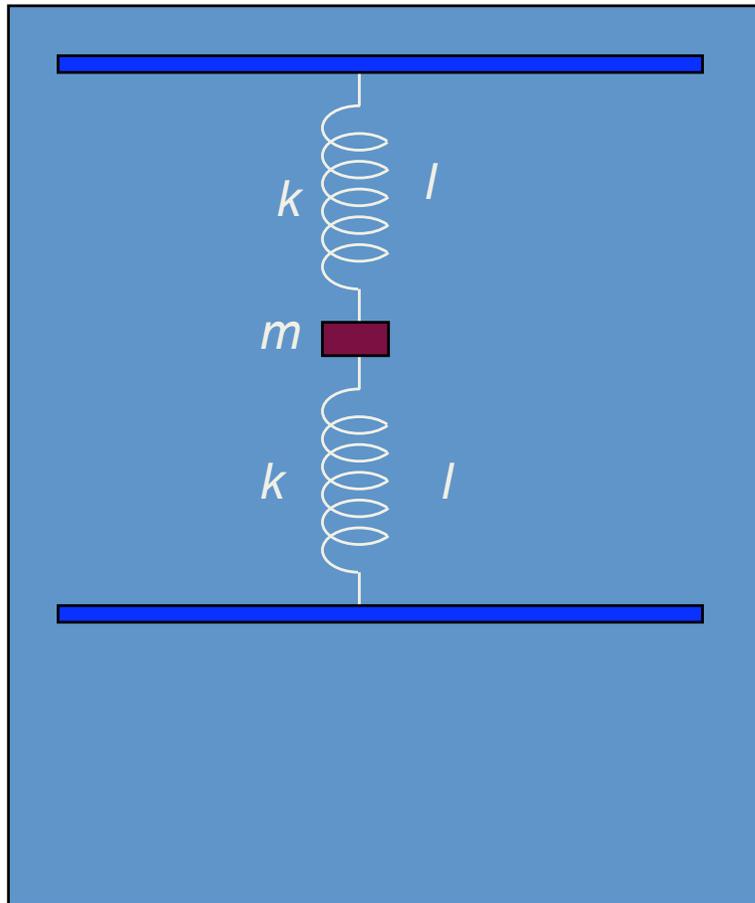
Let's look at an example of
non-linear oscillator:

the Duffing Oscillator

$$\ddot{x} + \delta\dot{x} + \beta x + \alpha x^3 = \gamma \cos \omega t$$

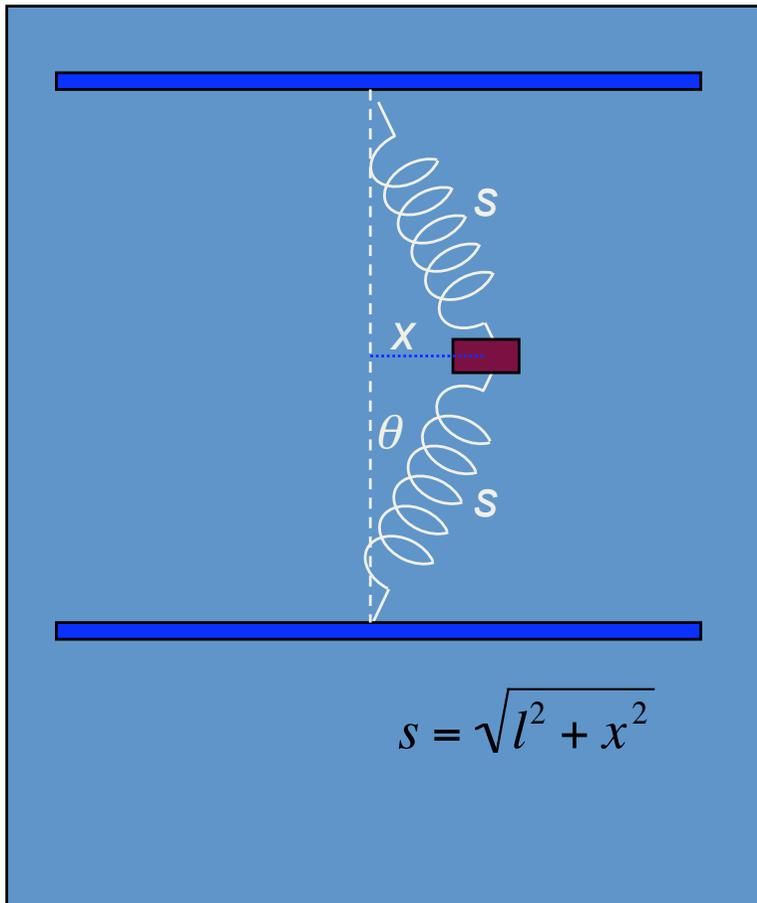
$$U(x) = \frac{1}{2}\beta x^2 + \frac{1}{4}\alpha x^4$$

A two springs system



- A mass is held between two springs.
 - Spring constant k
 - Natural length l
- Springs are on a horizontal surface.
 - Frictionless
 - No gravity

Transverse Displacement



- The force for a displacement is due to both springs.
 - Only transverse component
 - Looks like its harmonic

$$F = -2k\left(\sqrt{l^2 + x^2} - l\right)\sin\theta$$

$$= -2k\left(\sqrt{l^2 + x^2} - l\right)\frac{x}{\sqrt{l^2 + x^2}}$$

$$= -2kx\left(1 - \frac{1}{\sqrt{1 + x^2/l^2}}\right)$$

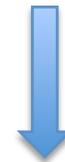
Purely Nonlinear

- The force can be expanded as a power series near equilibrium.
 - Expand in x/l

$$F = -2kl \frac{x}{l} \left(1 - \frac{1}{\sqrt{1 + x^2/l^2}} \right)$$

- The lowest order term is non-linear.

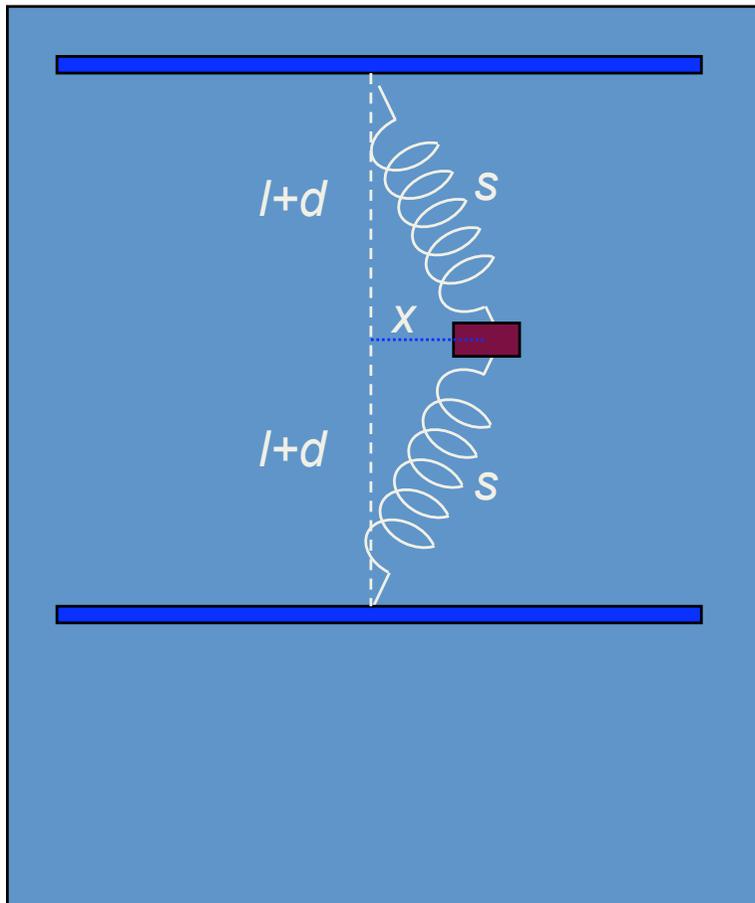
$$F \cong -kl \left(\frac{x}{l} \right)^3 + \dots$$



- Quartic potential
 - Not just a perturbation

$$V \cong \frac{k}{4l^2} x^4 + \dots$$

Mixed Potential



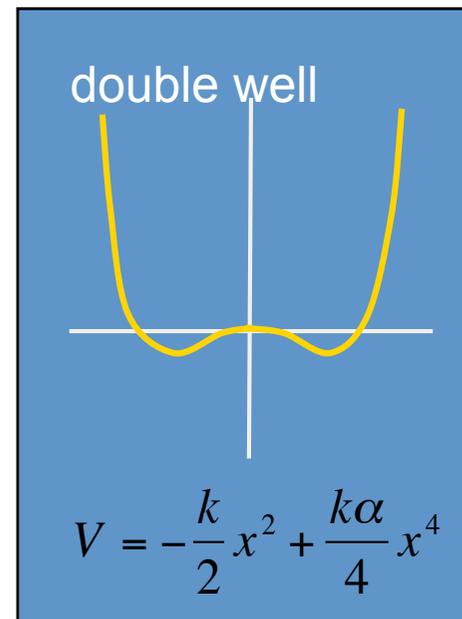
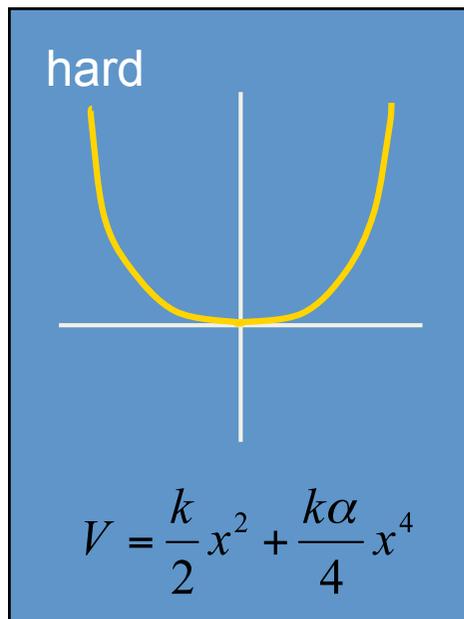
- Typical springs are not at natural length.
 - Approximation includes a linear term

$$F \cong -\frac{2kd}{l}x - \frac{k(l-d)}{l^3}x^3 + \dots$$

$$V \cong \frac{kd}{l}x^2 + \frac{k(l-d)}{4l^3}x^4 + \dots$$

Quartic Potentials

- The sign of the forces influence the shape of the potential.



Driven System

- Assume a more complete, realistic system.
 - Damping term
 - Driving force

$$m\ddot{x} = -\beta\dot{x} - kx - k\alpha x^3 + f \cos \omega t$$

$$\ddot{x} + \gamma\dot{x} + \omega_0^2 x + \alpha\omega_0^2 x^3 = f \cos \omega t$$

- Rescale the problem:
 - Set t such that $\omega_0^2 = k/m = 1$
 - Set x such that $k\alpha/m = 1$

- This is the Duffing equation

$$\ddot{x} + \gamma\dot{x} + x + x^3 = f \cos \omega t$$

Steady State Solution

- Try a solution, match terms

$$x(t) = A(\omega) \cos[\omega t - \theta(\omega)]$$

$$\ddot{x} + \gamma \dot{x} + x + x^3 = f \cos \omega t$$

$$A(1 - \omega^2) \cos(\omega t - \theta) - A\gamma\omega \sin(\omega t - \theta) + A^3 \cos^3(\omega t - \theta) = f \cos \omega t$$

trigonometric
identities

$$\cos^3(\omega t - \theta) = \frac{3}{4} \cos(\omega t - \theta) + \frac{1}{4} \cos 3(\omega t - \theta)$$

$$f \cos \omega t = f \cos \theta \cos(\omega t - \theta) - f \sin \theta \sin(\omega t - \theta)$$

$$[A(1 - \omega^2 + \frac{3}{4} A^2) - f \cos \omega t] \cos(\omega t - \theta)$$

$$+ [-A\gamma\omega + f \sin \omega t] \sin(\omega t - \theta)$$

$$+ \frac{1}{4} A^3 \cos 3(\omega t - \theta)$$

$$= 0$$

$$f \cos \omega t = A(1 - \omega^2 + \frac{3}{4} A^2)$$

$$f \sin \omega t = A\gamma\omega$$

$$\frac{1}{4} A^3 \cos 3(\omega t - \theta) \approx 0$$

Amplitude Dependence

- Find the amplitude-frequency relationship.
 - Reduces to forced harmonic oscillator for $A \rightarrow 0$

$$f^2 = A^2[(1 - \omega^2)^2 + (\gamma\omega)^2]$$

- Find the case for minimal damping and driving force.
 - f, γ both near zero
 - Defines resonance condition

$$f^2 \cos^2 \omega t = A^2(1 - \omega^2 + \frac{3}{4} A^2)^2$$

$$f^2 \sin^2 \omega t = A^2 \gamma^2 \omega^2$$

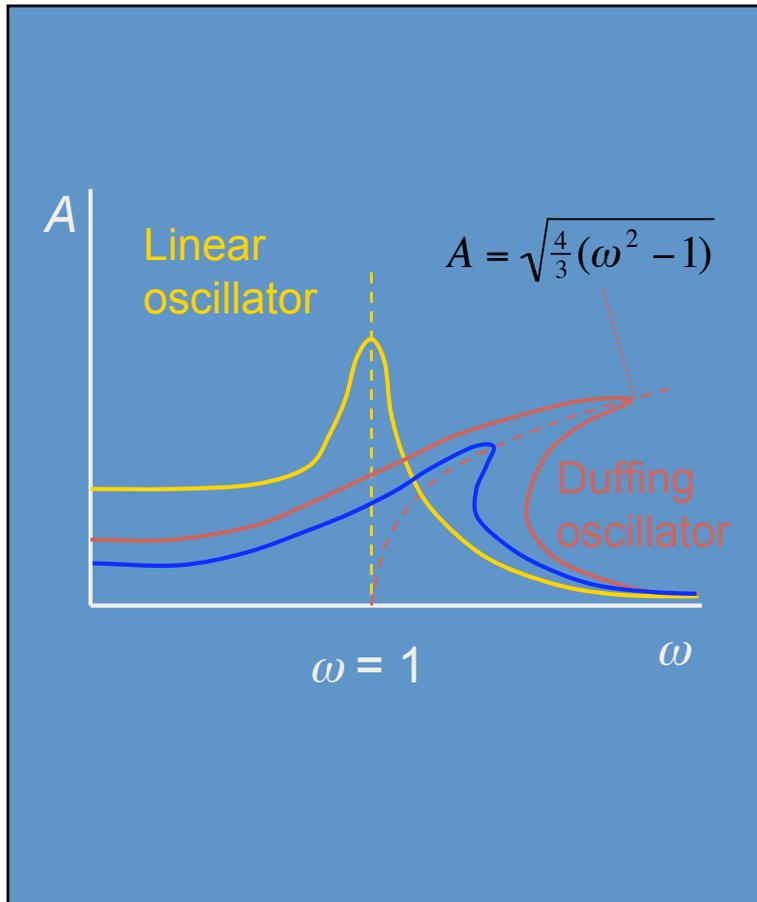
$$f^2 = A^2[(1 - \omega^2 + \frac{3}{4} A^2)^2 + \gamma^2 \omega^2]$$

$$0 = A^2[(1 - \omega^2 + \frac{3}{4} A^2)^2 + 0]$$

$$0 = 1 - \omega^2 + \frac{3}{4} A^2$$

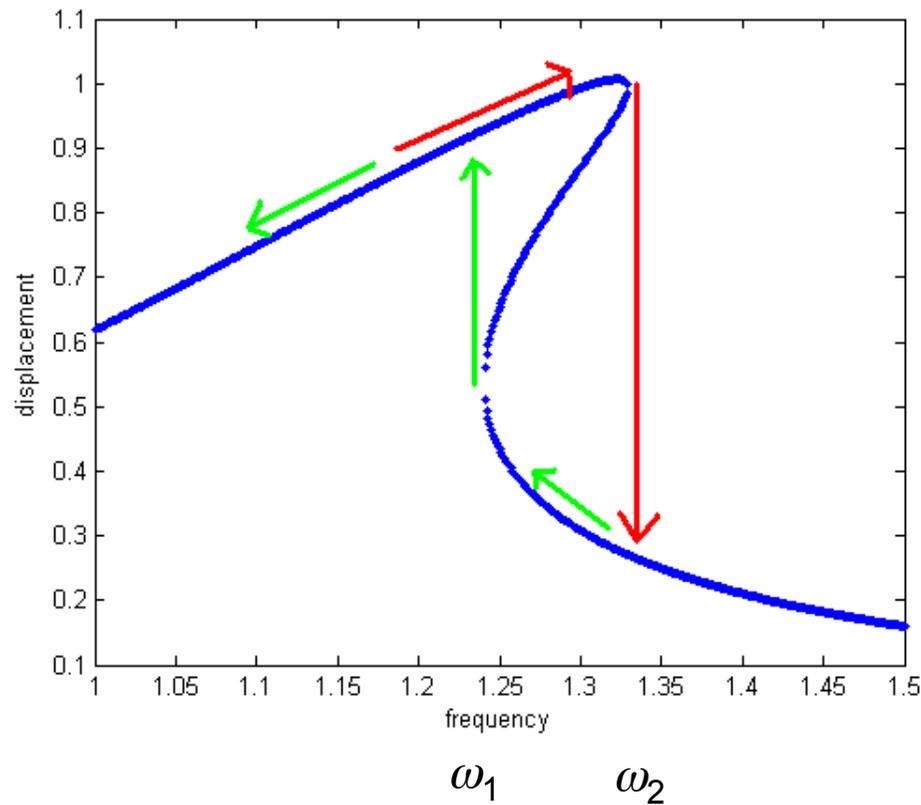
$$A(\omega) = \sqrt{\frac{4}{3}(\omega^2 - 1)}$$

Nonlinear Resonance Frequency



- The resonance frequency of a linear oscillator is independent of amplitude.
- The resonance frequency of a **Duffing oscillator** increases with amplitude.

... brings to hysteresis



- A Duffing oscillator behaves differently for increasing and decreasing frequencies.
 - Increasing frequency has a jump in amplitude at ω_2
 - Decreasing frequency has a jump in amplitude at ω_1
- This is hysteresis.

Nonlinear Resonance

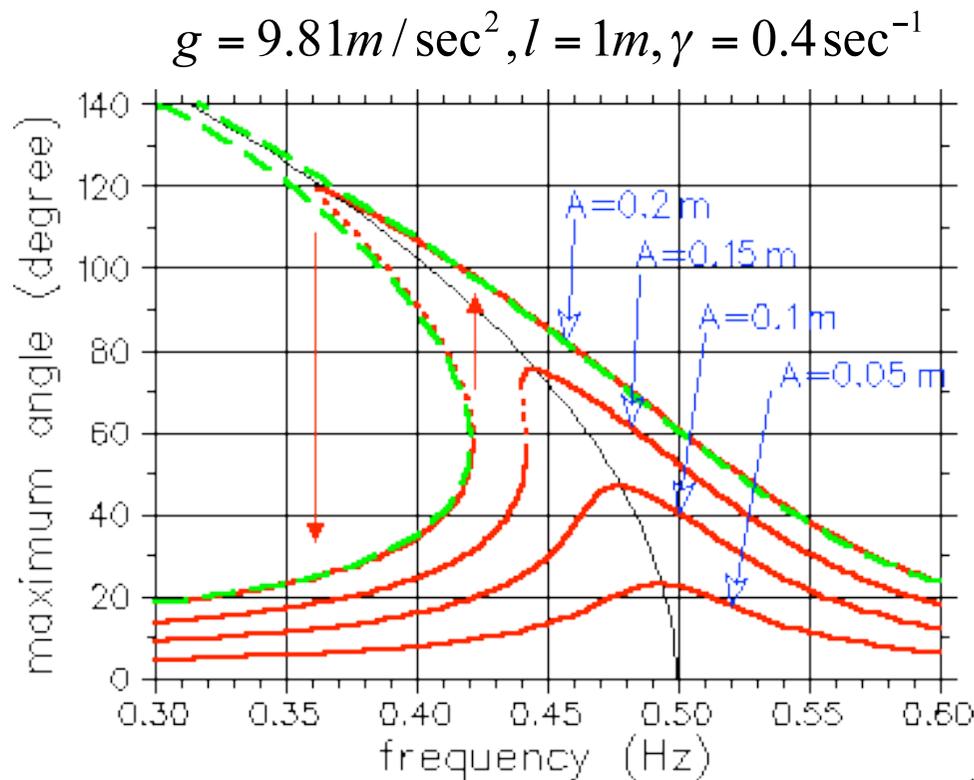
(in general...)

Nonlinear resonance seems not to be so much different from the (linear) resonance of a harmonic oscillator. But both, the dependency of the eigenfrequency of a nonlinear oscillator on the amplitude and the nonharmonicity of the oscillation lead to a behavior that is impossible in harmonic oscillators, namely, the **foldover effect** and **superharmonic resonance**, respectively.

Both effects are especially important in the case of weak damping.

The foldover effect

The **foldover** effect got its name from the bending of the resonance peak in a amplitude versus frequency plot. This bending is due to the frequency-amplitude relation which is typical for nonlinear oscillators.



Foldover effect for a pendulum

The pendulum eq.:

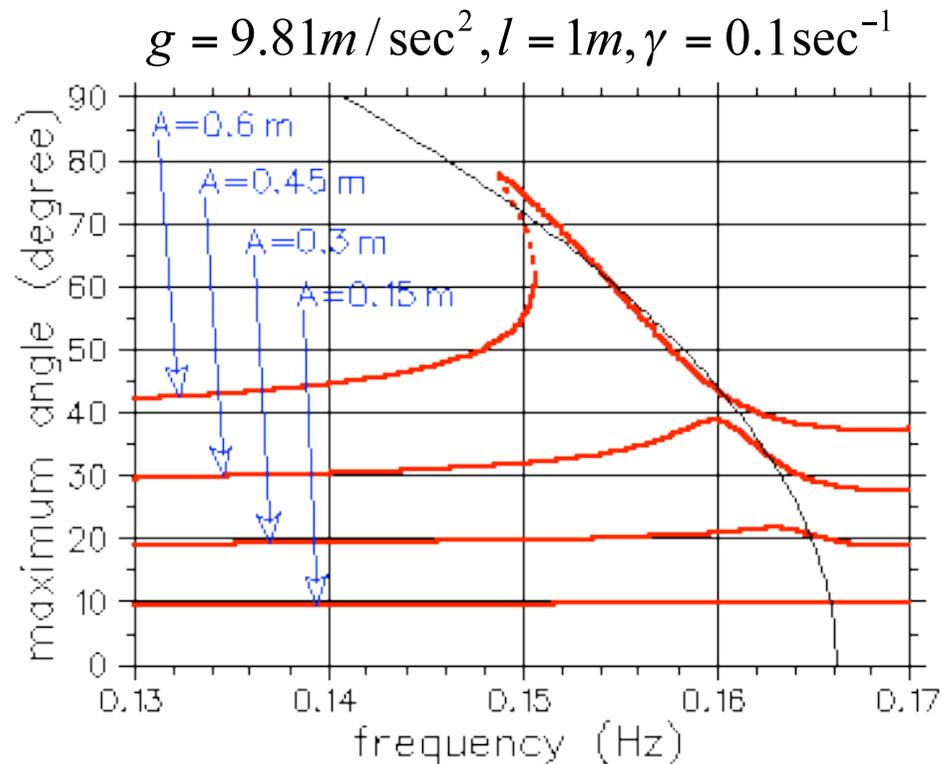
$$\ddot{\varphi} = -\gamma\dot{\varphi} - \omega_0^2 \sin \varphi + f \cos \omega t$$

$$\omega_0^2 = \frac{g}{l}$$

The superharmonic resonance

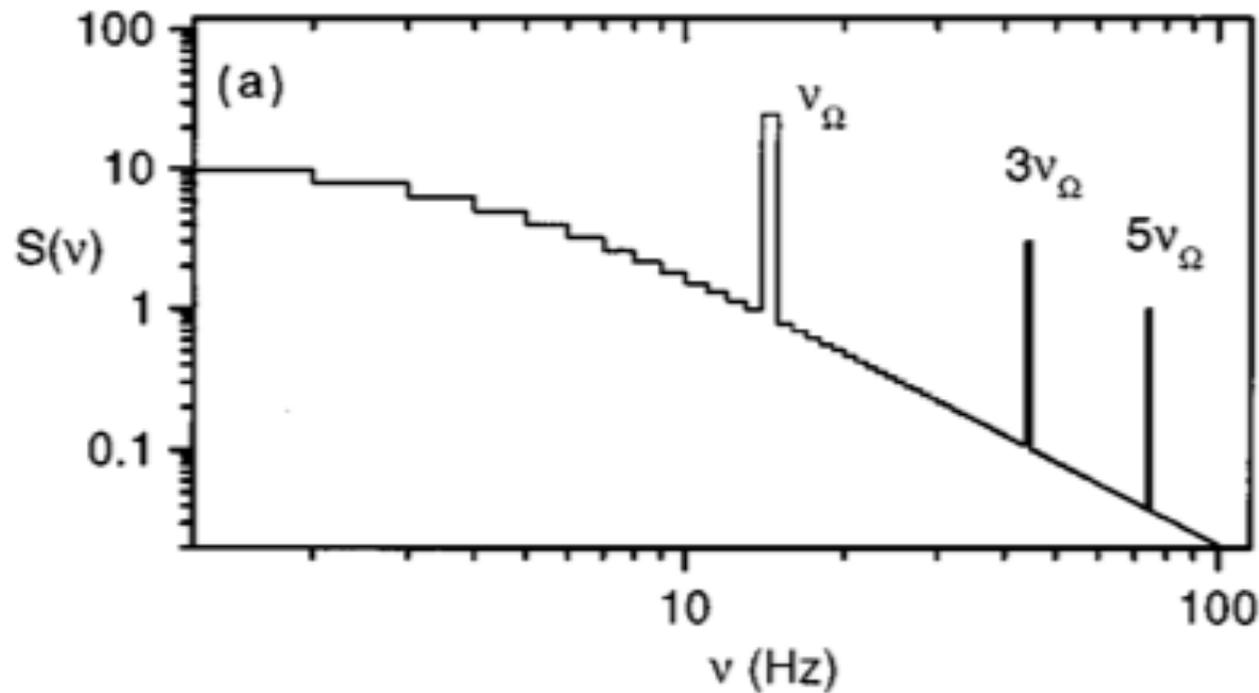
Nonlinear oscillators do not oscillate sinusoidal.

Superharmonic resonance is simply the resonance with one of this higher harmonics of a nonlinear oscillation. In an amplitude/frequency plot appear additional resonance peaks. In general, they appear at driving frequencies which are integer fractions of the fundamental frequency.



Bistable Duffing

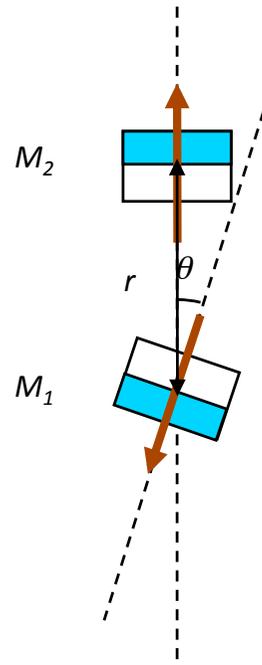
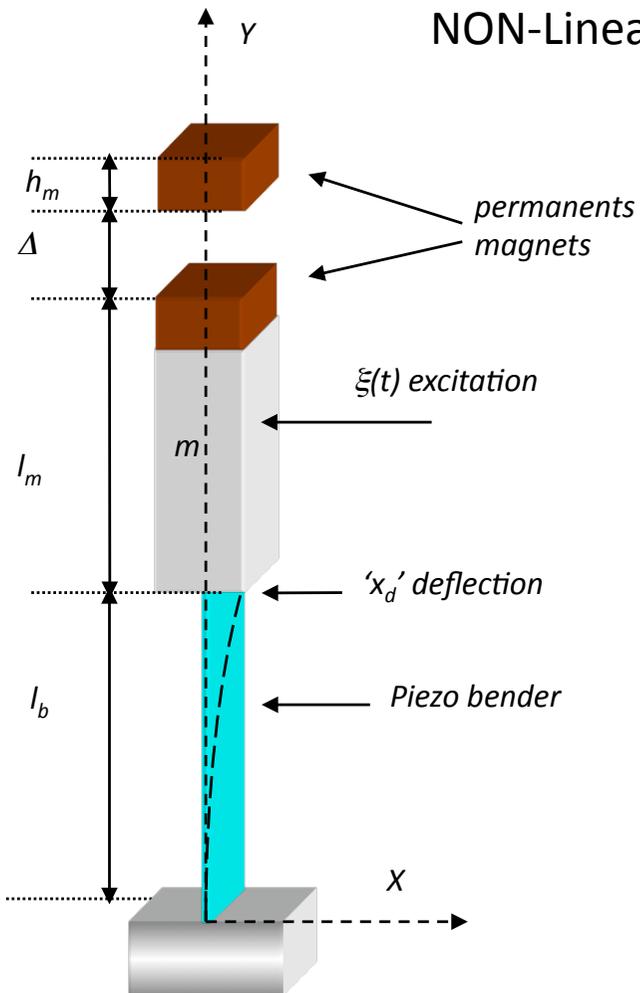
In case of a bistable oscillator the frequency response for an overdamped system is highly spread in the low frequency region.



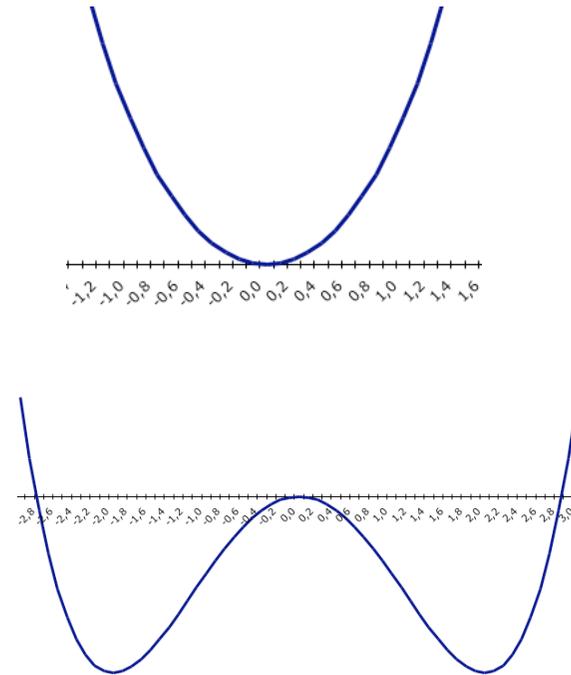
Noise energy harvesting

NON-Linear mechanical oscillators

NON-Linear Inverted pendulum

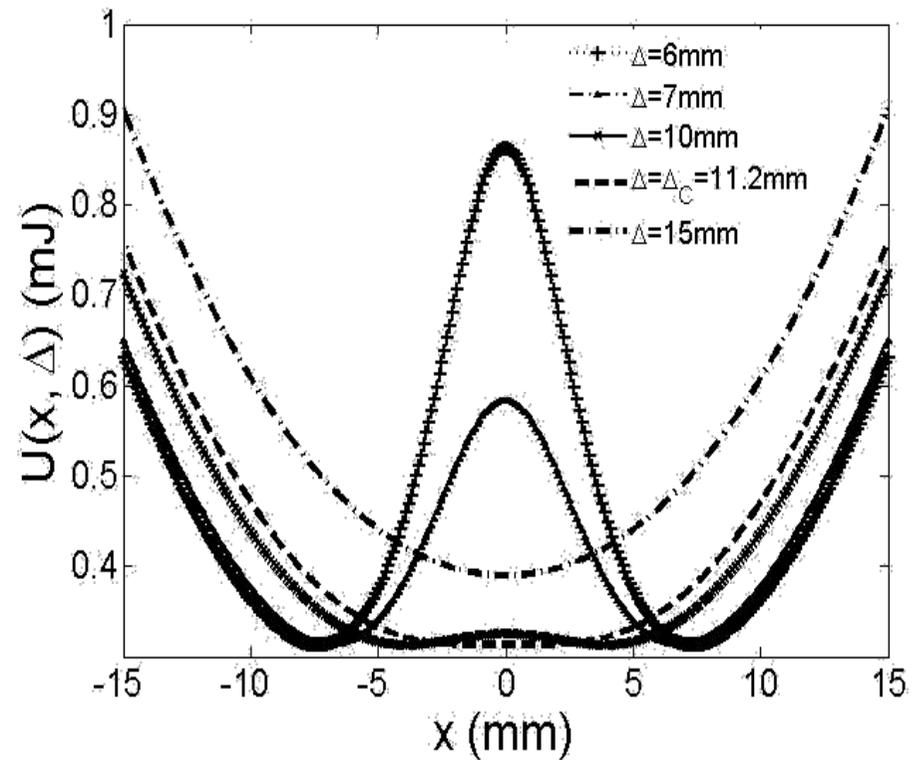
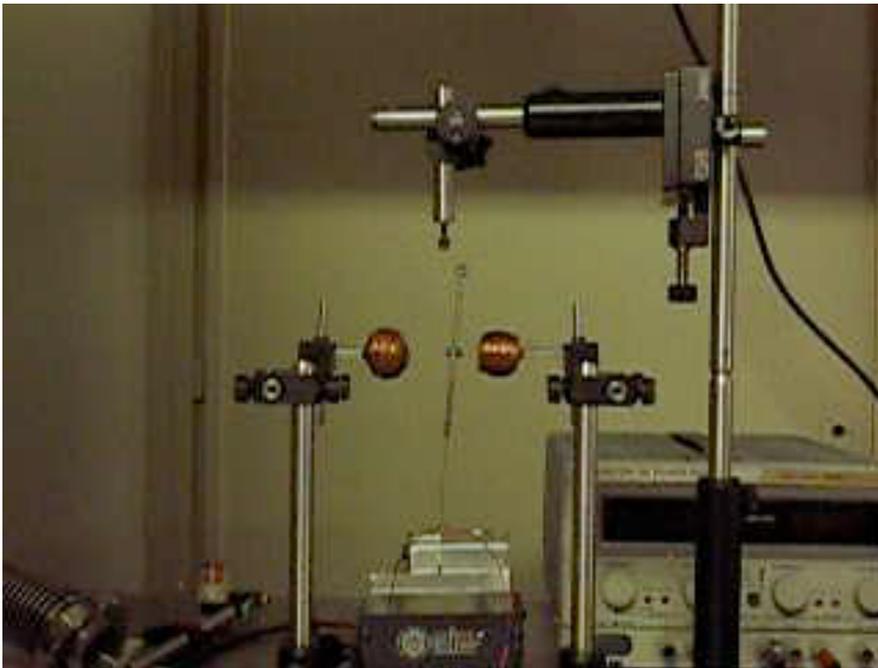


b)



Noise energy harvesting

NON-Linear mechanical oscillators



<http://www.nipslab.org/node/1676>

Nonlinear Energy Harvesting, F. Cottone; H. Vocca; L. Gammaitoni
Physical Review Letters, 102, 080601 (2009)

Noise energy harvesting

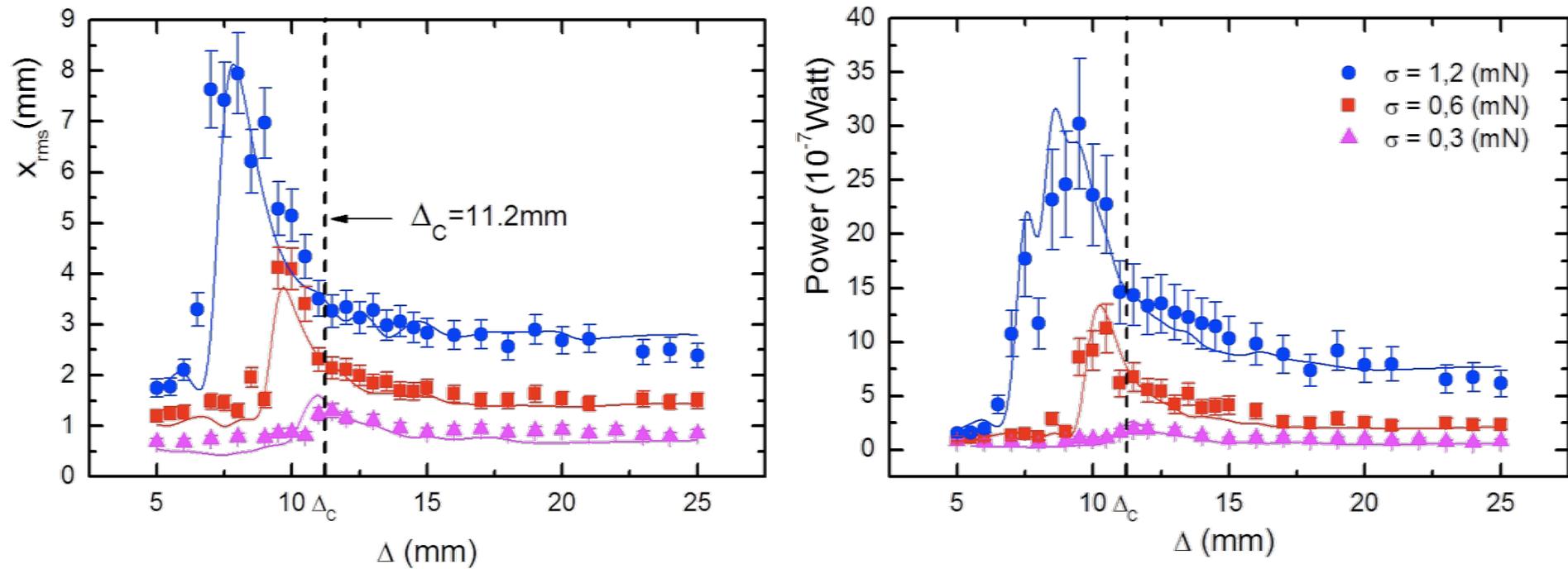
NON-Linear mechanical oscillators



Nonlinear Energy Harvesting, F. Cottone; H. Vocca; L. Gammaitoni
Physical Review Letters, 102, 080601 (2009)

Noise energy harvesting

NON-Linear mechanical oscillators

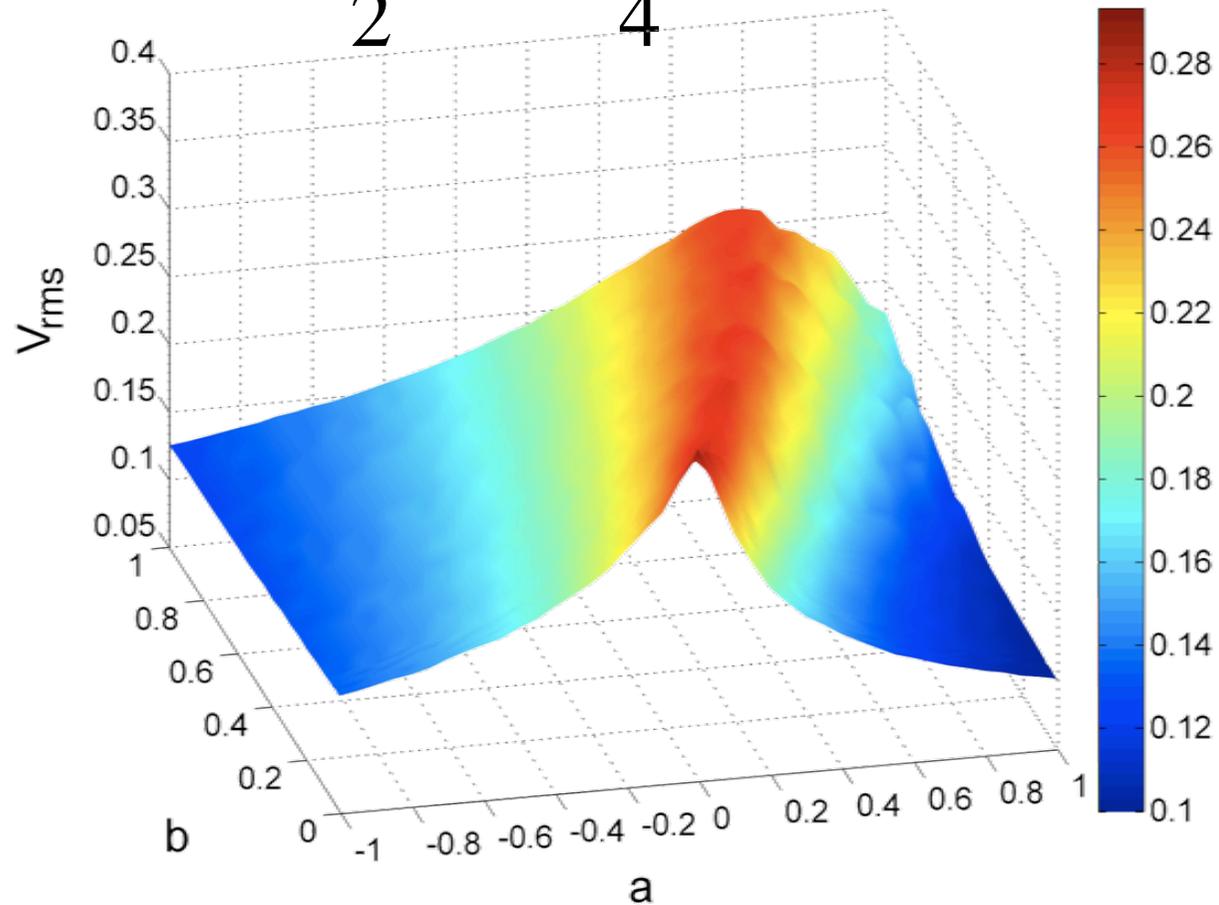


Noise energy harvesting

Non-linear systems

$$U(x) = -\frac{1}{2}ax^2 + \frac{1}{4}bx^4$$

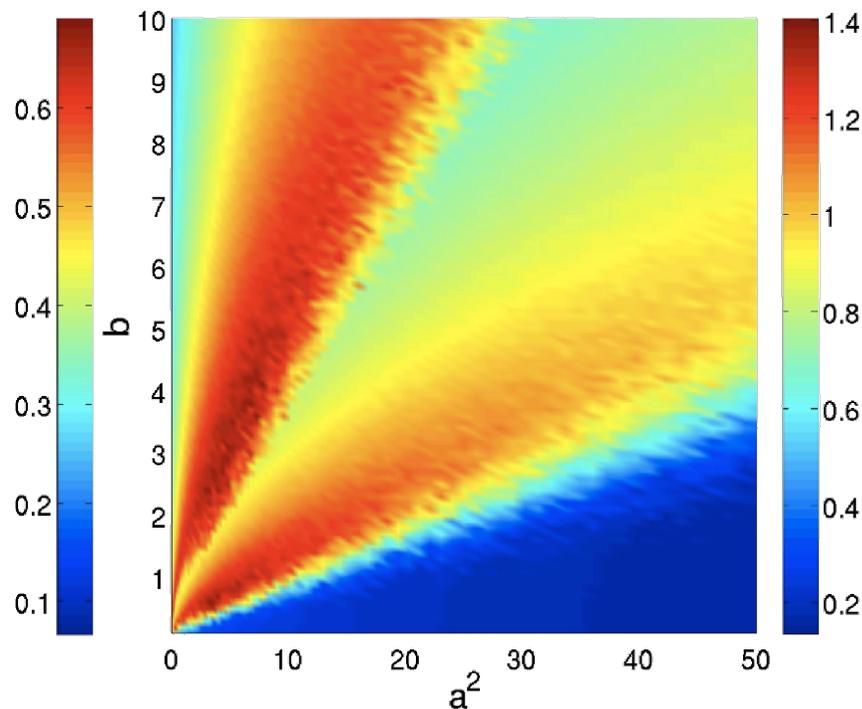
Duffing potential



Noise energy harvesting

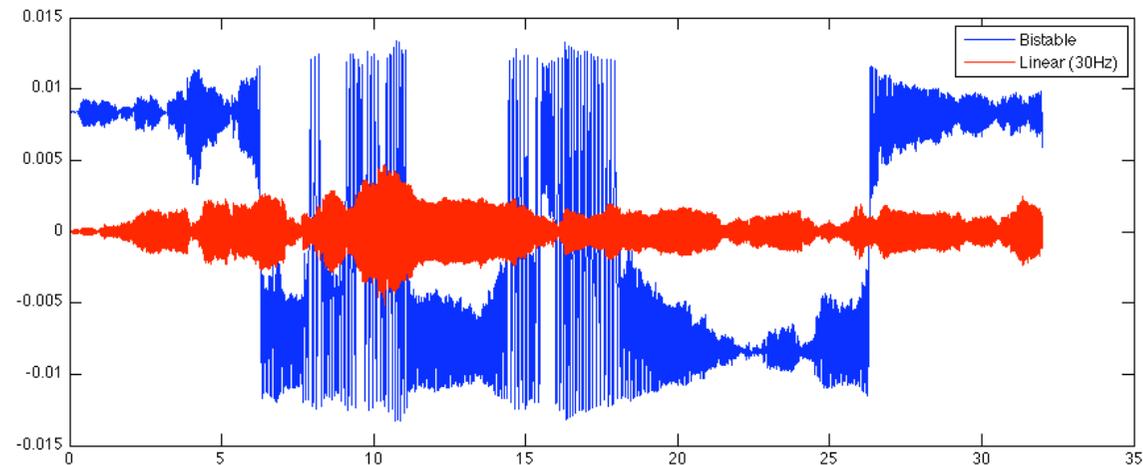
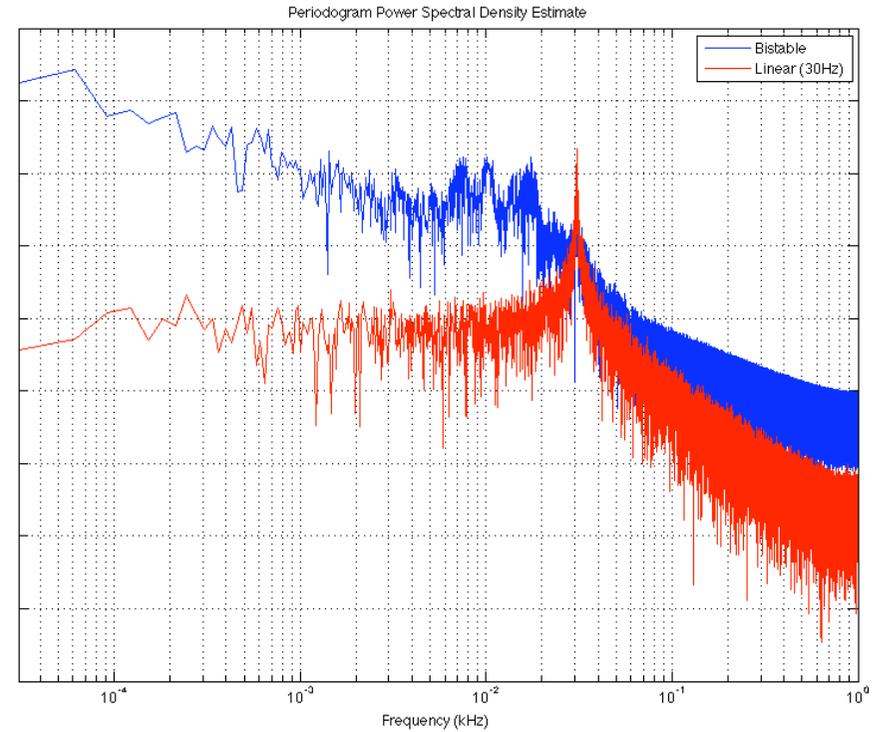
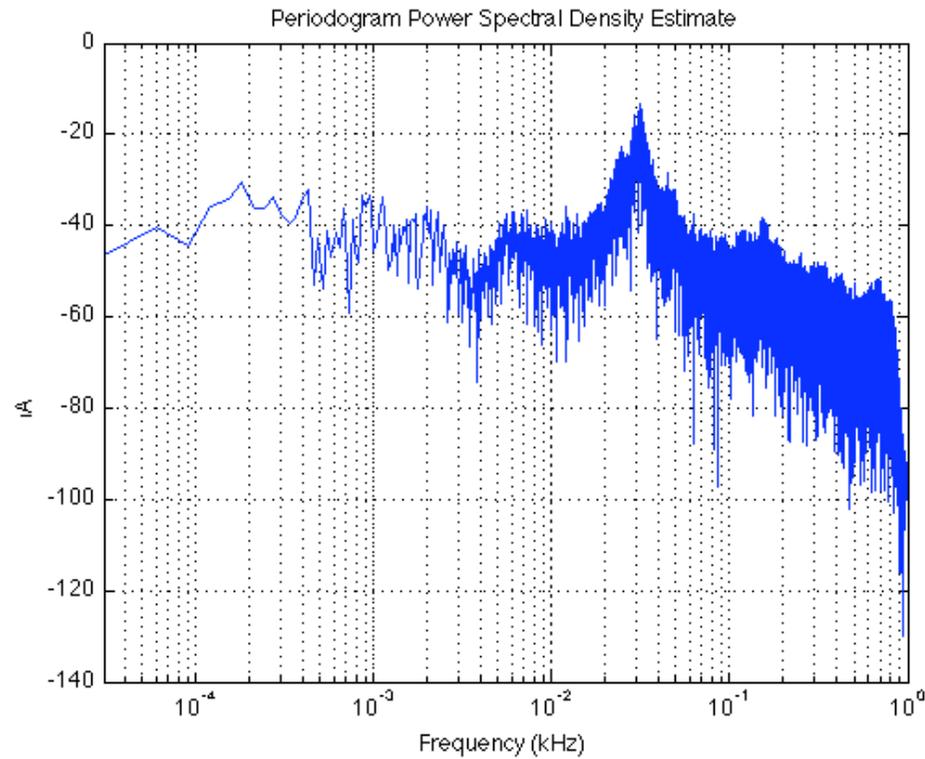
Non-linear systems

$$U(x) = -\frac{1}{2}ax^2 + \frac{1}{4}bx^4 \quad \text{Duffing potential}$$

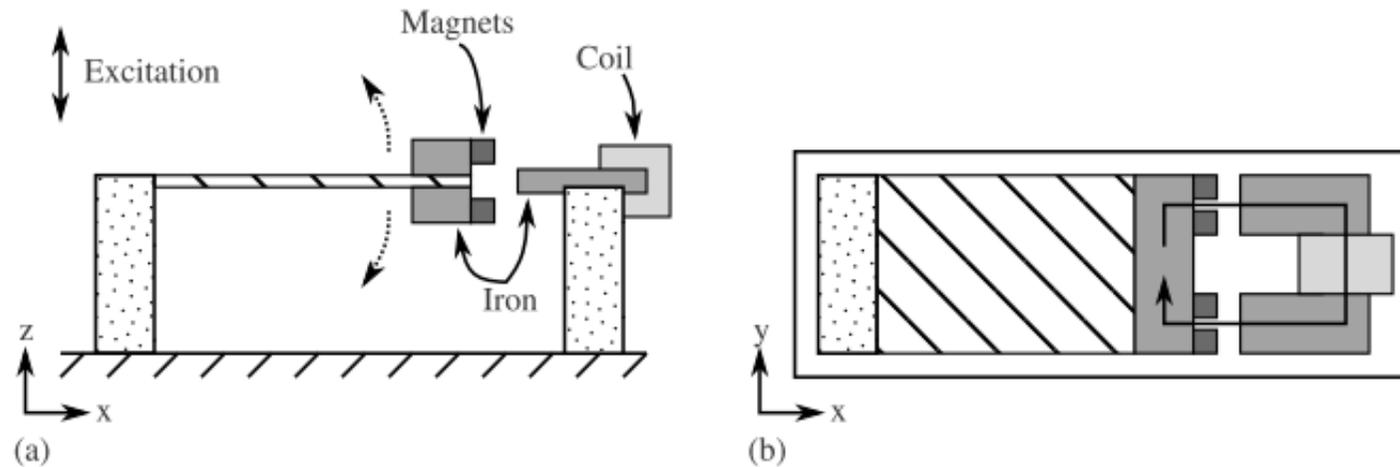


$$b_{MAX} = \frac{a^2}{4D \log(\tau_p)}$$

A compared response

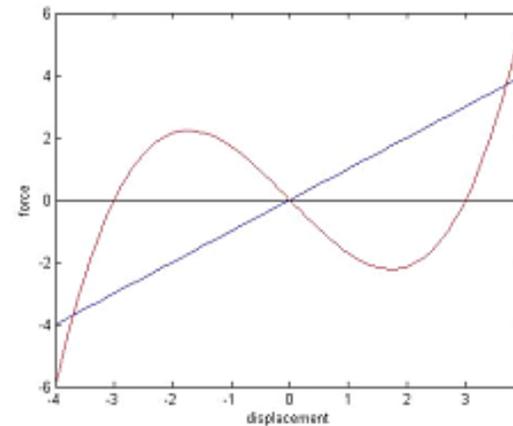
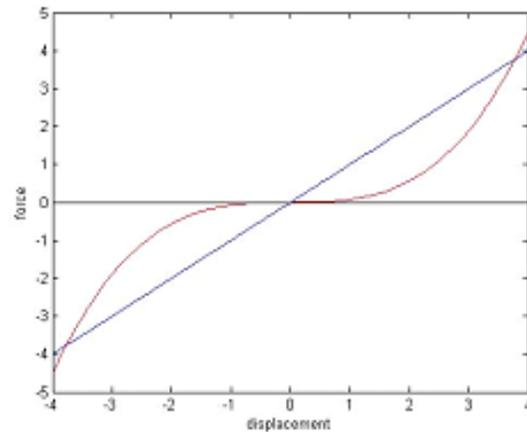


Barton D.A.W., Burrow S.G. and Clare L.R., 2010, "Energy Harvesting from Vibrations with a Nonlinear Oscillator," *Journal of Vibration and Acoustics*, 132, 021009.



$$m\ddot{x} + \left(c + \frac{\theta^2}{R_C + R_L}\right)\dot{x} + kx + \beta x^3 = F \sin(\omega t)$$

$k < 0 \implies$ hardening
 $k > 0 \implies$ bistable



After non-dimensionalization: $\ddot{x} + 2\xi_{eff}\dot{x} + x + \beta x^3 = \Gamma \sin(\omega t)$

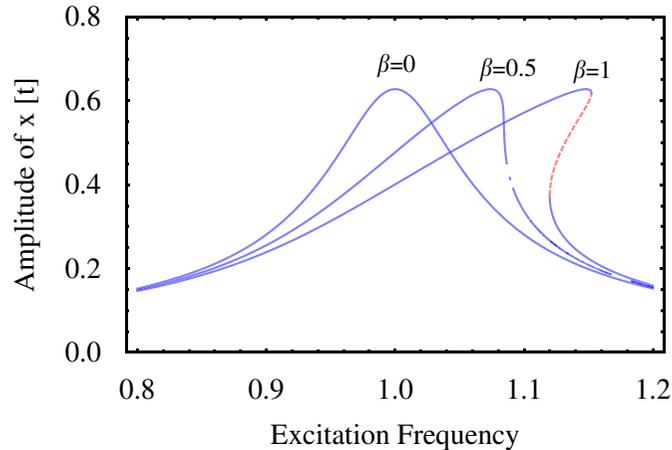
Noise energy harvesting

Only bistability???

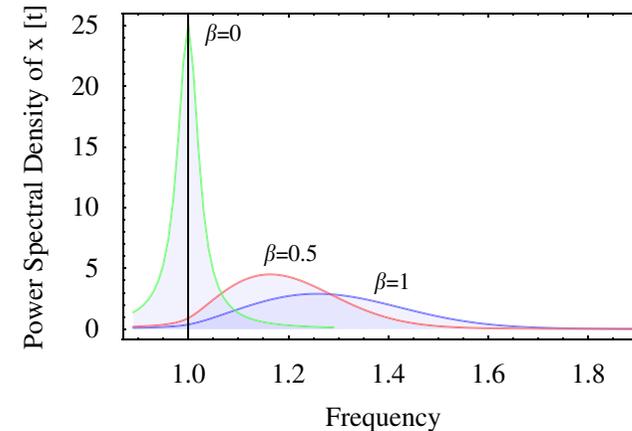
Considering a Duffing oscillator: $\ddot{x} + 2\xi_{eff}\dot{x} + x + \beta x^3 = F(t)$

ξ_{eff} is the effective damping ratio for both electrical and mechanical damping

$\beta > 0$ is a stiffness nonlinearity coefficient



Steady-state frequency response curves under harmonic excitations of a fixed frequency



Power spectral density curves of $x(t)$ under White Gaussian excitations of a fixed spectral density.

Noise energy harvesting

For an electromagnetic duffing-type harvester, (similar to the one shown before):

$$\ddot{u} + 2\xi\omega_n\dot{u} + \omega_n^2u + \theta i + \beta u^3 = F(t)$$

$$\theta i = iR$$

under a Gaussian White noise :

$$\langle \xi(t)\xi(t_1) \rangle = 2\pi\sigma\delta(t - t_1)$$

u represents the position of the mass, ξ is the mechanical damping ratio, ω_n is the natural frequency, β is a cubic nonlinearity coefficient, θ is an electromechanical coupling coefficient, R is the load resistance, and i is the current passing through the load.

The expected value of the power (mean power) is: $\langle P \rangle = \frac{\theta\sigma}{\xi_{eff}}$ Not dependent on the nonlinear coefficient

For small nonlinearities (small β) and various “colored” noises the mean output power decreases as the nonlinearity increases indicating that the nonlinearity does not improve the output power even when the excitations are Colored.

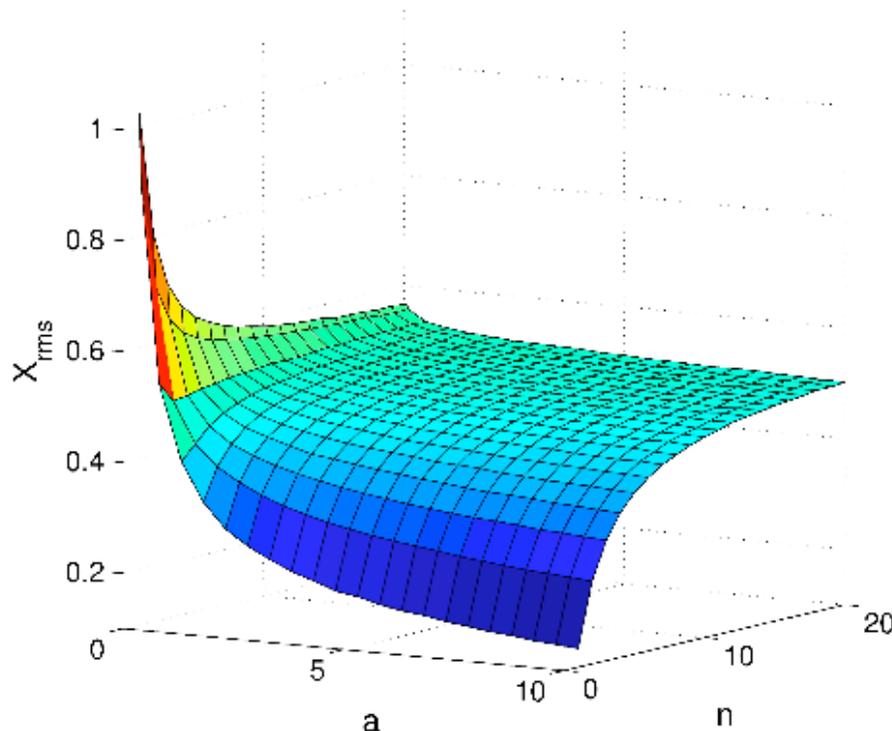
More details on: M.F. Daqaq, Journal of Sound and Vibration 329 (2010) 3621–3631

Noise energy harvesting

Only bistability???

A more general monostable potential...

$$U(x) = ax^{2n} \quad \text{with} \quad \begin{aligned} a &> 0 \\ n &= 1, 2, \dots \end{aligned}$$



In an exponentially correlated noise with correlation time τ :

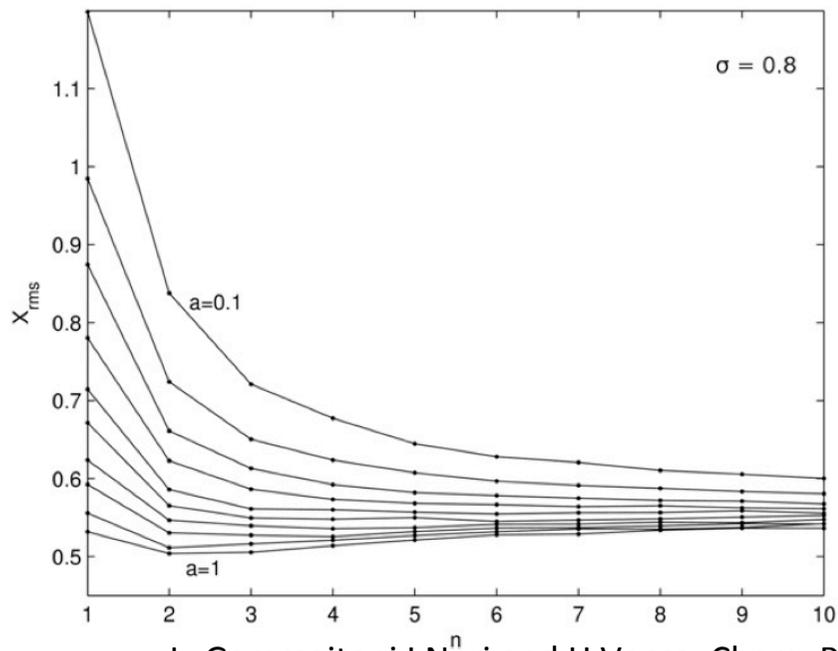
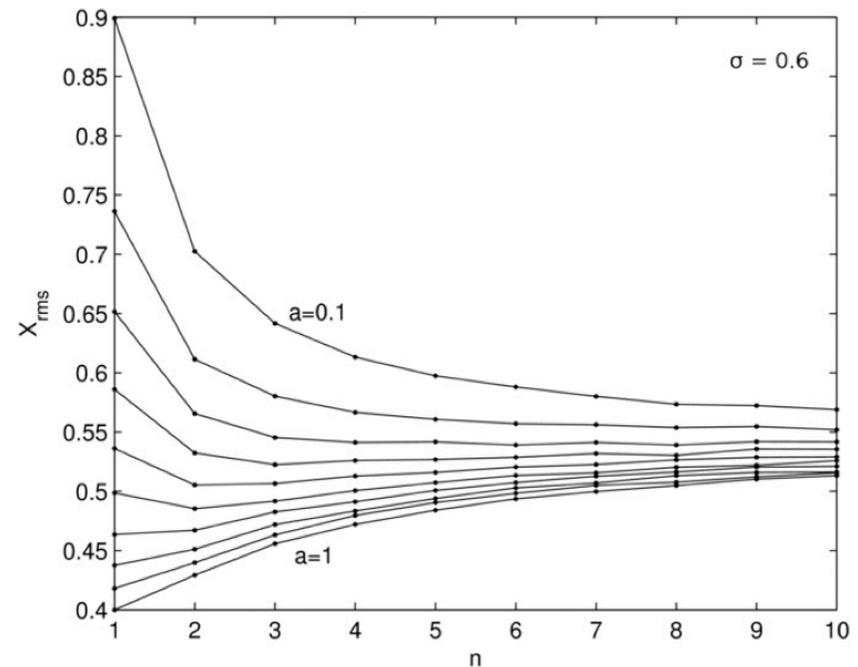
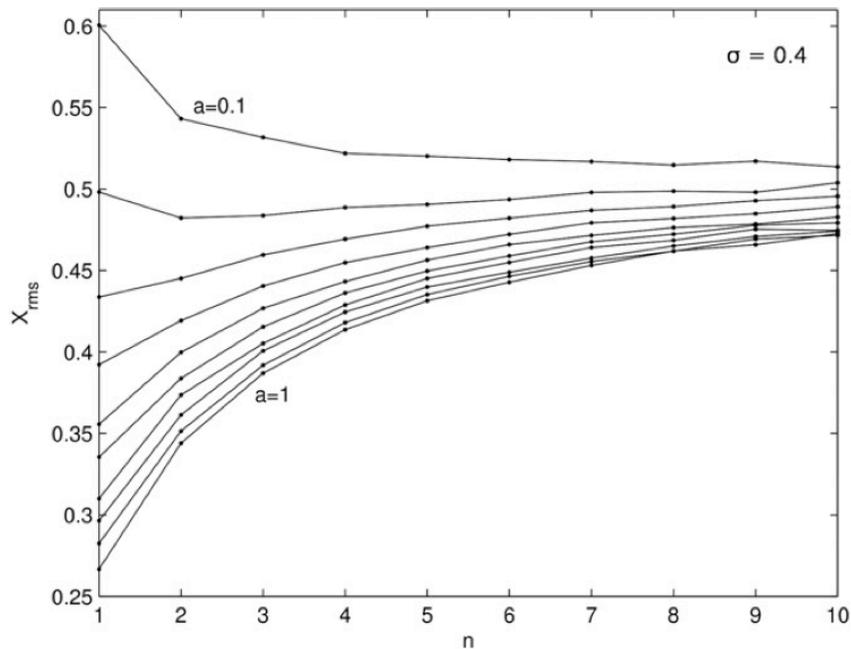
$$\langle \xi(t)\xi(t_1) \rangle = \sigma^2 e^{-\frac{|t-t_1|}{\tau}}$$

There exists a threshold amplitude a_{th} :

$$a_{th} \approx \frac{D}{4} = \sigma^2 \tau$$

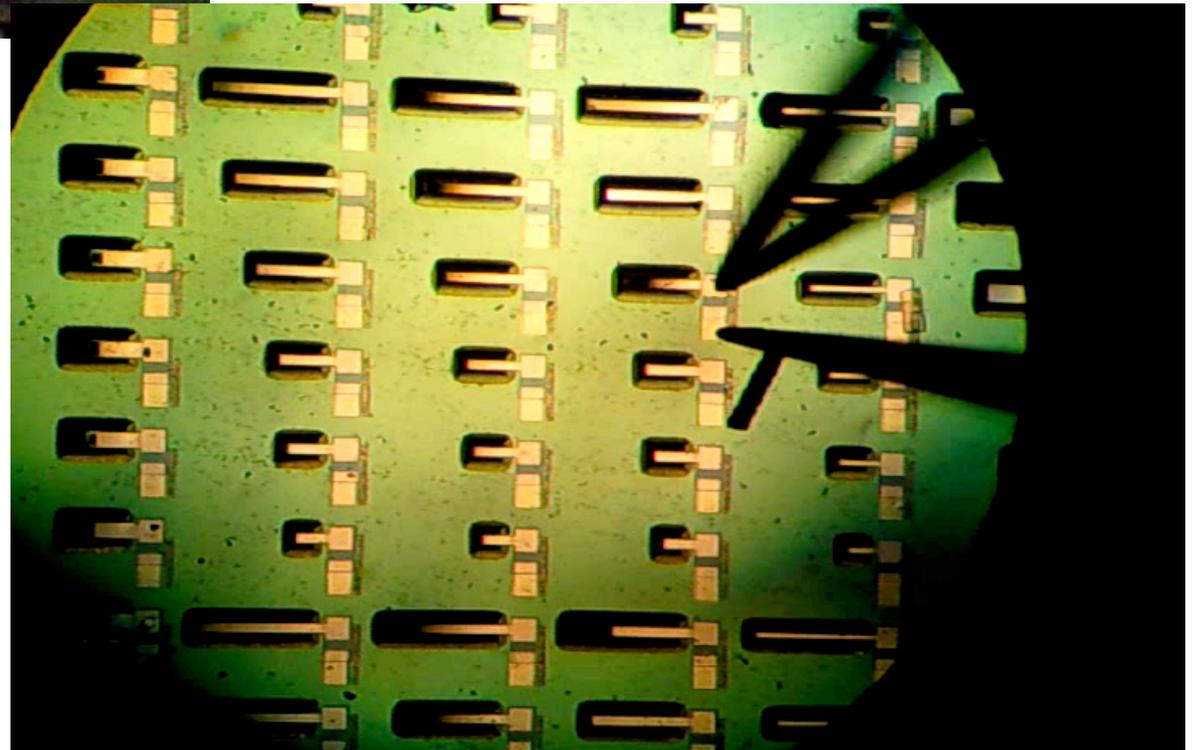
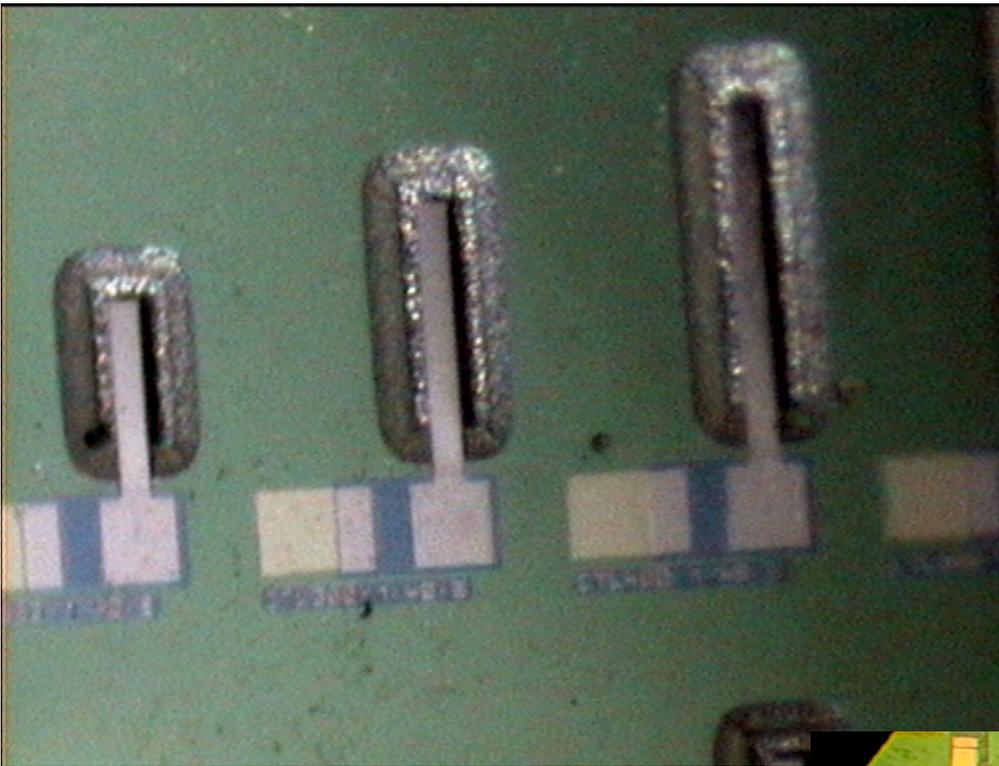
Above which the nonlinear system outperforms the linear one.

Varying the noise amplitude

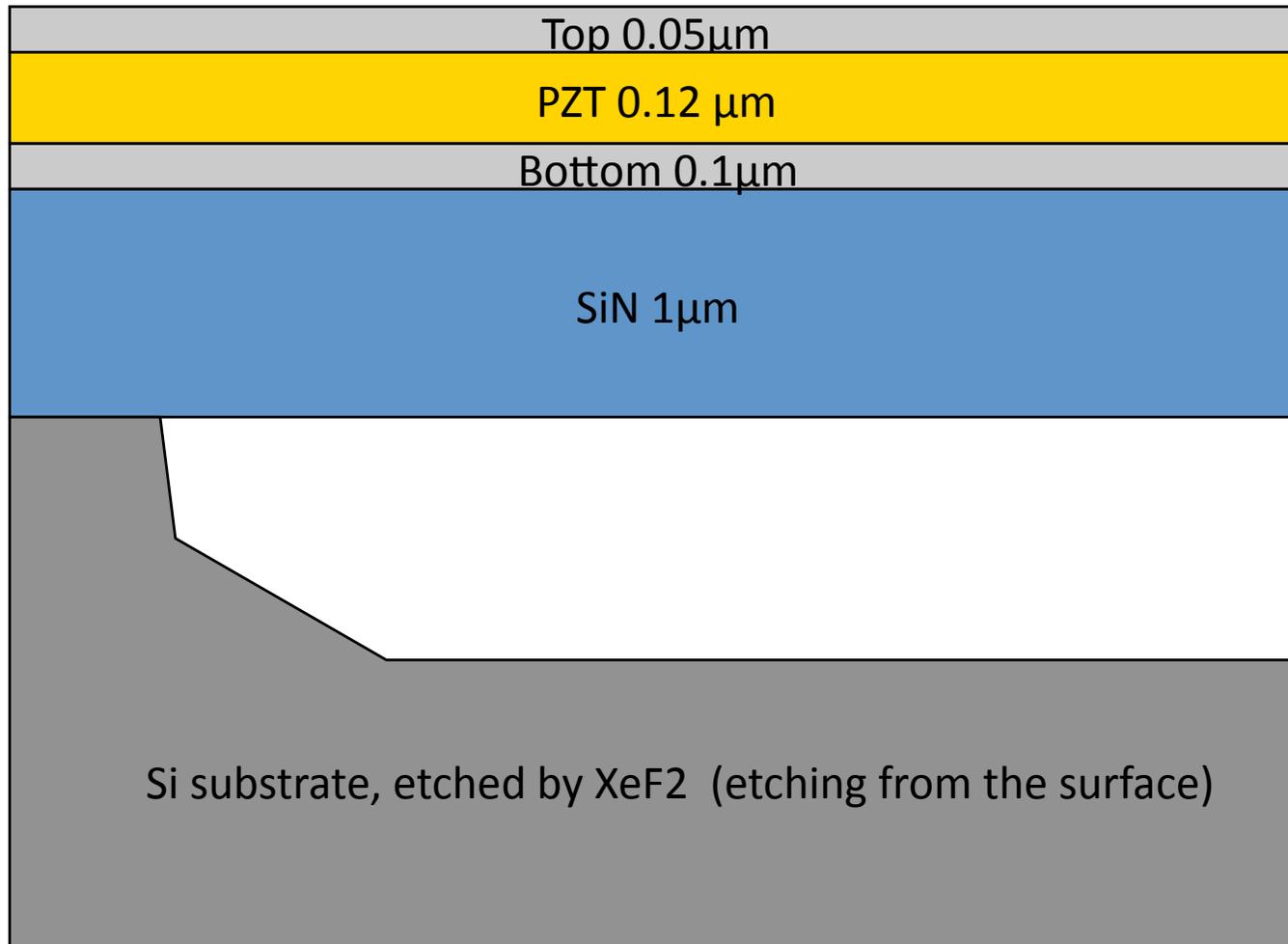


Once σ and a are fixed the choice of a linear ($n = 1$) or nonlinear potential ($n \geq 2$) can be made in order to maximize X_{rms} and consequently the power obtained at the device output.

In collaboration with
CEA-Leti we are
investigating
 μ cantilevers
dynamics



Scheme of the cross section of one cantilever



Typical electrical features

- Max voltage sustainable: 5-7V
- e_{31} PZT = -5C/m^2
- Dielectric constant PZT: 1000

Conclusions

- 1) Non resonant (i.e. non-linear) mechanical oscillators can outperform resonant (i.e. linear) ones
- 2) Non-linear systems are more difficult to treat but more interesting...
- 3) Bistability is not the only nonlinearity available...
- 4) The same principles are also valid for capacitive and inductive harvesters
- 5) A great amount of work has still to be done... good for us!!!

A macro bistable application has been developed (by wisepower srl)...

